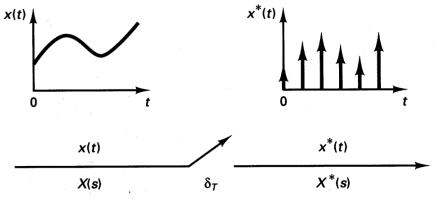
Discrete-Time Control Systems

Introduction

- Backgrounds necessary for the analysis and design of discrete-time control systems in the *z* plane are presented.
 - The main advantage of the *z* transform method: it enables us to apply conventional continuous-time design methods to discretetime systems.
- The chapter covers:
 - Mathematical representation of the sampling operation
 - The convolution integral method for obtaining the z transform
 - The sampling theorem based on the fact that the Laplace transform of the sampled signal is periodic
 - Mathematical modeling of digital controllers in terms of pulse transfer function
 - Realization of digital controllers and digital filters

- Impulse Sampling
 - A fictitious sampler
 - The output of the sampler is a train of impulses.



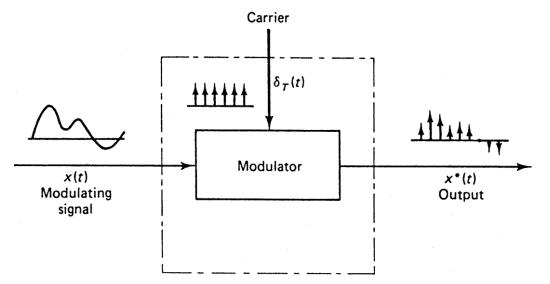
$$x^{*}(t) = \sum_{k=0}^{\infty} x(kT)\delta(t - kT)$$

$$x^*(t) = x(0)\delta(t) + x(T)\delta(t-T) + \dots + x(kT)\delta(t-kT) + \dots$$

- Impulse Sampling (cont.)
 - Let's define a train of unit impulses:

$$\delta_T(t) = \sum_{k=0}^{\infty} \delta(t - kT)$$

- The sampler may be considered a modulator with
 - The modulating signal: the input *x*(*t*)
 - The carrier : the train of unit impulses $\delta_T(t)$



- Impulse Sampling (cont.)
 - The Laplace transform of $x^*(t)$

$$X^{*}(s) = L[x^{*}(t)] = x(0)L[\delta(t)] + x(T)L[\delta(t-T)] + x(2T)L[\delta(t-2T)] + \cdots$$
$$= x(0) + x(T)e^{-Ts} + x(2T)e^{-2Ts} + \cdots$$
$$= \sum_{k=0}^{\infty} x(kT)e^{-kTs}$$
If we define $e^{Ts} = z$ or $s = \frac{1}{T}\ln z$

$$X^{*}(s)\Big|_{s=(1/T)\ln z} = X^{*}\left(\frac{1}{T}\ln z\right) = \sum_{k=0}^{\infty} x(kT)z^{-k} = X(z)$$

- Hence we may write

$$X^*(s)\Big|_{s=(1/T)\ln z} = X(z)$$

- Impulse Sampling (cont.)
 - Summary
 - If the continuous-time signal *x*(*t*) is impulse sampled in a periodic manner, the sampled signal may be represented by

$$x^*(t) = \sum_{k=0}^{\infty} x(t)\delta(t - kT)$$

• The Laplace transform of the impulse-sampled signal *x*^{*}(*t*) has been shown to be the same as the *z* transform of signal *x*(*t*) if

$$e^{Ts}=z$$

- Data-Hold Circuits
 - Data-hold: a process of generating a continuous-time signal h(t) from a discrete-time sequence x(kT).
 - A hold circuit approximately reproduces the signal applied to the sampler.

$$h(kT + \tau) = a_n \tau^n + a_{n-1} \tau^{n-1} + \dots + a_1 \tau + a_0$$
 where $0 \le \tau < T$

Note that signal h(kT) must equal x(kT):

h(kT) = x(kT)

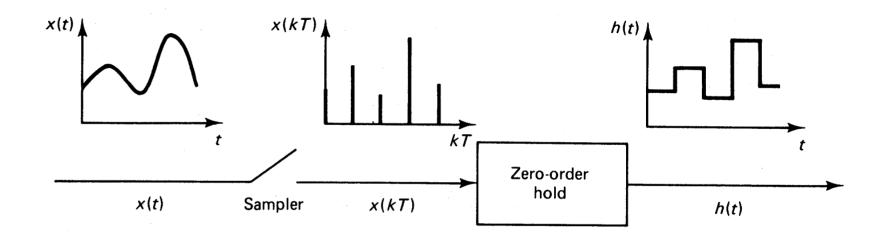
 $\implies h(kT+\tau) = a_n \tau^n + a_{n-1} \tau^{n-1} + \dots + a_1 \tau + x(kT) \qquad \text{n-th order hold}$

- The simplest data-hold: zero-order hold (clamper)

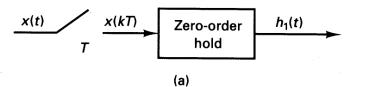
$$h(kT + \tau) = x(kT)$$

• Zero-Order Hold

 $h(kT+t) = x(kT), \quad \text{for } 0 \le t < T$



Zero-Order Hold (cont.)



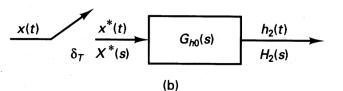
A real sampler and zero-order hold

$$h_{1}(t) = x(0) [1(t) - 1(t - T)] + x(T) [1(t - T) - 1(t - 2T)] + x(2T) [1(t - 2T) - 1(t - 3T)] + \cdots$$
$$= \sum_{k=0}^{\infty} x(kT) [1(t - kT) - 1(t - (k + 1)T)]$$

Since
$$L[1(t-kT)] = \frac{e^{-kTs}}{s}$$

 $\Longrightarrow L[h_1(t)] = H_1(s) = \sum_{k=0}^{\infty} x(kT) \frac{e^{-kTs} - e^{-(k+1)Ts}}{s} = \frac{1 - e^{-Ts}}{s} \sum_{k=0}^{\infty} x(kT) e^{-kTs}$

Zero-Order Hold (cont.)



Mathematical model: an impulse sampler and transfer function

$$L[h_{2}(t)] = H_{2}(s) = H_{1}(s)$$
$$H_{2}(s) = \frac{1 - e^{-Ts}}{s} \sum_{k=0}^{\infty} x(kT) e^{-kTs}$$

From the figure $H_2(s) = G_{h0}(s)X^*(s)$

Since
$$X^*(s) = \sum_{k=0}^{\infty} x(kT)e^{-kTs}$$

$$\implies H_2(s) = \frac{1 - e^{-Ts}}{s} X^*(s)$$

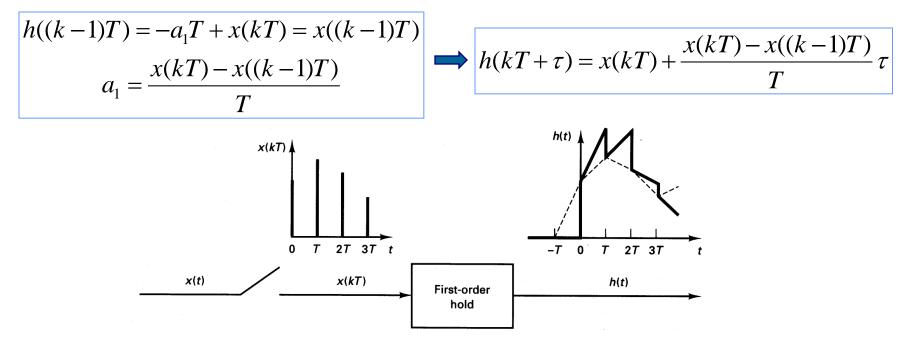
$$\implies G_{h0}(s) = \frac{1 - e^{-Ts}}{s}$$

Transfer function of First-Order Hold

 $h(kT+\tau) = a_n \tau^n + a_{n-1} \tau^{n-1} + \dots + a_1 \tau + x(kT) \quad \text{n-th order hold}$

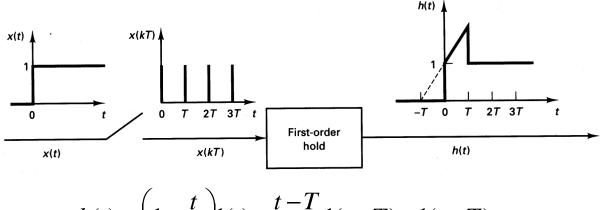
1-st order hold $h(kT+\tau) = a_1\tau + x(kT)$

By applying the condition that h((k-1)T) = x((k-1)T)



Transfer function of First-Order Hold (cont.)

Derivation of the transfer function



 $h(t) = \left(1 + \frac{t}{T}\right) l(t) - \frac{t - T}{T} l(t - T) - l(t - T)$

Taking the Laplace transform

$$H(s) = \left(\frac{1}{s} + \frac{1}{Ts^2}\right) - \frac{1}{Ts^2}e^{-Ts} - \frac{1}{s}e^{-Ts}$$
$$= \frac{1 - e^{-Ts}}{s} + \frac{1 - e^{-Ts}}{Ts^2} = \left(1 - e^{-Ts}\right)\frac{Ts + 1}{Ts^2}$$