

Digital Transmission Via Carrier Modulation

Carrier Phase Modulation

BPSK(Binary PSK) : $M=2$

QPSK(Quadrature PSK): $M=4$

MPSK(M-ary PSK): $M > 2$

Carrier Phase Modulation

- Carrier phase is used to transmit digital information via digital phase modulation

- M-ary phase modulation ($M=2^k$)

- $\theta_m = \frac{2\pi m}{M}, \quad m = 0, 1, \dots, M-1$

- Binary phase modulation ($M=2$)

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- M carrier phase modulated signal waveform

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$$u_m(t) = A g_T(t) \cos(2\pi f_c t + \frac{2\pi m}{M}), \quad m = 0, 1, \dots, M-1$$

↑
↑
Transmitting filter pulse shape
Amplitude

PSK(Phase Shift Keying)

- Digital phase modulation is called PSK
- PSK signals have equal energy

- See Figure 7.7 page 289

- $\mathfrak{S}_m = \int_{-\infty}^{\infty} u_m^2(t) dt = \frac{A^2}{2} \int_{-\infty}^{\infty} g_T^2(t) dt \equiv \mathfrak{S}_s, \text{ for all } m$

- Normalized power case (Rectangular $g_T(t)$)

- $g_T(t) = \sqrt{2/T}, 0 \leq t \leq T \quad \frac{1}{2} \int_{-\infty}^{\infty} g_T^2(t) dt = 1 \quad A = \sqrt{\mathfrak{S}_s}$

- $u_m(t) = \sqrt{\frac{2\mathfrak{S}_s}{T}} \cos(2\pi f_c t + \frac{2\pi m}{M}), \quad m = 0, 1, \dots, M-1$

↑ Constant envelope

PSK with 2 orthogonal basis

- Changing representation of waveform

$$\begin{aligned}
 u_m(t) &= A g_T(t) \cos(2\pi f_c t + \frac{2\pi m}{M}) \\
 &= \sqrt{S_s} g_T(t) \left\{ \cos(2\pi f_c t) \cos(\frac{2\pi m}{M}) - \sin(2\pi f_c t) \sin(\frac{2\pi m}{M}) \right\} \\
 &= \underbrace{\sqrt{S_s} \cos(\frac{2\pi m}{M})}_{S_{mc}} \underbrace{g_T(t) \cos(2\pi f_c t)}_{\mathcal{E}_1(t)} + \underbrace{\sqrt{S_s} \sin(\frac{2\pi m}{M})}_{S_{ms}} \underbrace{-g_T(t) \sin(2\pi f_c t)}_{\mathcal{E}_2(t)}
 \end{aligned}$$

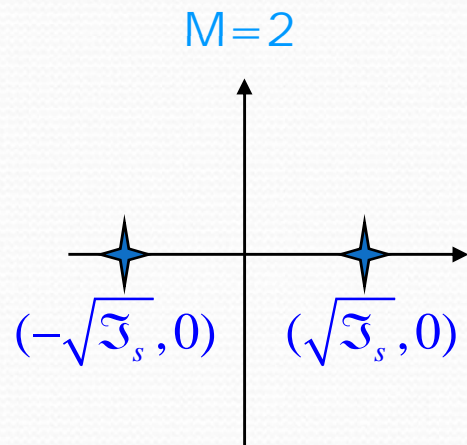
$\mathcal{E}_1(t), \mathcal{E}_2(t)$

- Two orthogonal basis:
- PSK can be represented geometrically as two-dimensional with components S_{mc}, S_{ms}

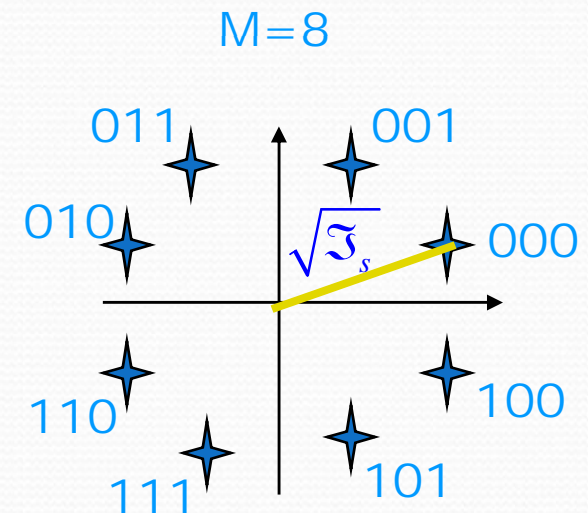
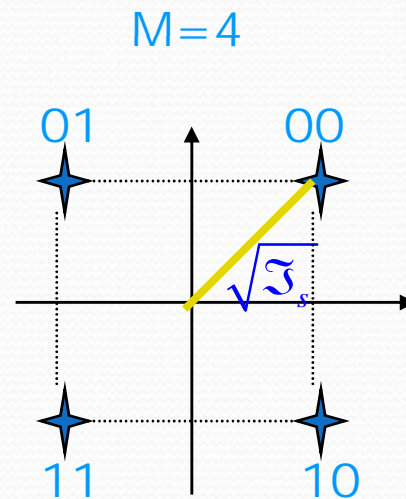
Constellation of PSK

- Two dimensional

- $$s_m = (s_{mc}, s_{ms}) = \left(\sqrt{\mathcal{E}_s} \cos \frac{2\pi f m}{M}, \sqrt{\mathcal{E}_s} \sin \frac{2\pi f m}{M} \right)$$



Same as binary PAM
with antipodal signals



Why Gray code is used ?

Phase demodulation and detection

- Received signal through AWGN channel

- $$r(t) = u_m(t) + n(t)$$

- $$= s_{mc} g_T(t) \cos(2f_c t) - s_{ms} g_T(t) \sin(2f_c t)$$

- $$+ n_c(t) \cos(2f_c t) - n_s(t) \sin(2f_c t)$$

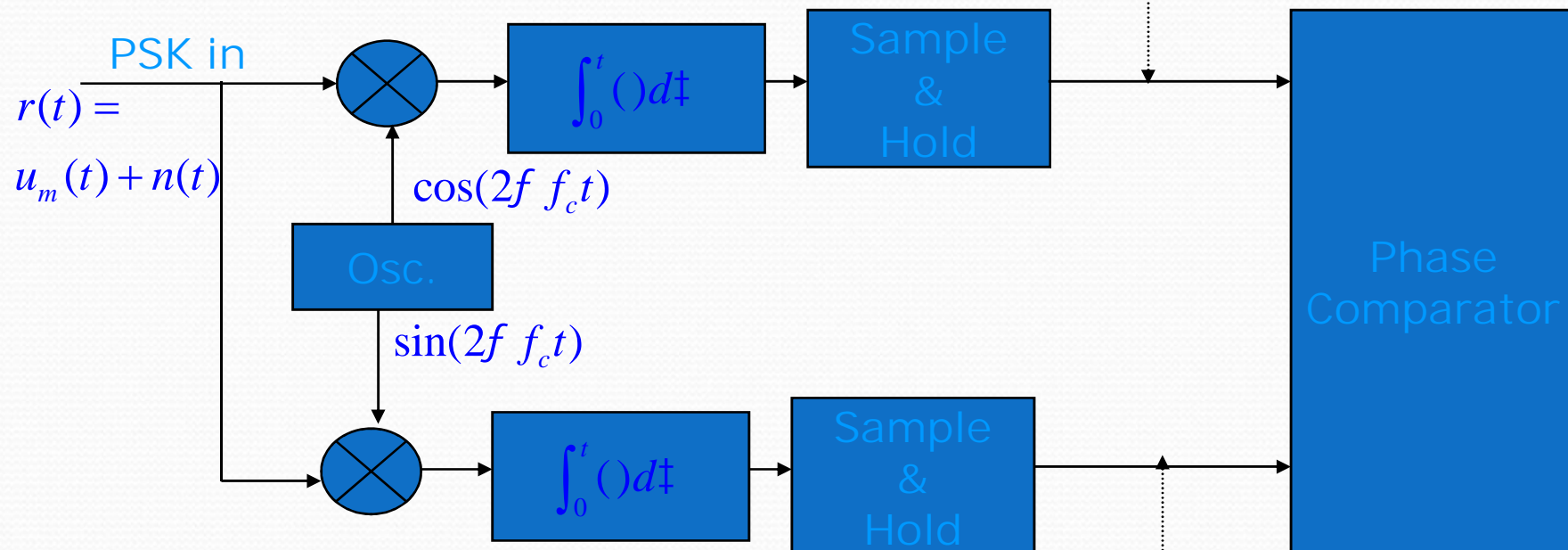
- Optimal Receiver

- Two correlator (or Matched Filter)
 - Two dimensional Signal
 - Optimum Detector

Optimum Receiver for PSK

- Two Correlator

$$r_c = \sqrt{\mathfrak{I}_s} \cos \frac{2f_m}{M} + \frac{1}{2} \int_{-\infty}^{\infty} g_T(t) n_c(t) dt = \sqrt{\mathfrak{I}_s} \cos \frac{2f_m}{M} + n_c$$



$$r_s = \sqrt{\mathfrak{I}_s} \sin \frac{2f_m}{M} + \frac{1}{2} \int_{-\infty}^{\infty} g_T(t) n_s(t) dt = \sqrt{\mathfrak{I}_s} \sin \frac{2f_m}{M} + n_s$$

Noise component

- Because $n_c(t)$ and $n_s(t)$ are uncorrelated zero mean Gaussian process

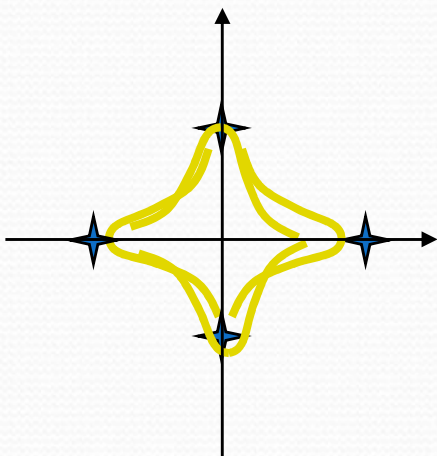
- Zero mean:

$$E[n_c] = E[n_s] = E[n_c n_s] = 0$$

- With variance:

$$E[n_c^2] = E[n_s^2] = \frac{N_0}{2}$$

- Example of $M=4$



$$r = (r_c, r_s) = \left(\sqrt{\mathcal{I}_s} \cos \frac{2f m}{M} + n_c, \sqrt{\mathcal{I}_s} \sin \frac{2f m}{M} + n_c \right)$$

$$r_0 = (\sqrt{\mathcal{I}_s} + n_c, n_c)$$

$$r_1 = (n_c, \sqrt{\mathcal{I}_s} + n_c)$$

$$r_2 = (-\sqrt{\mathcal{I}_s} + n_c, n_c)$$

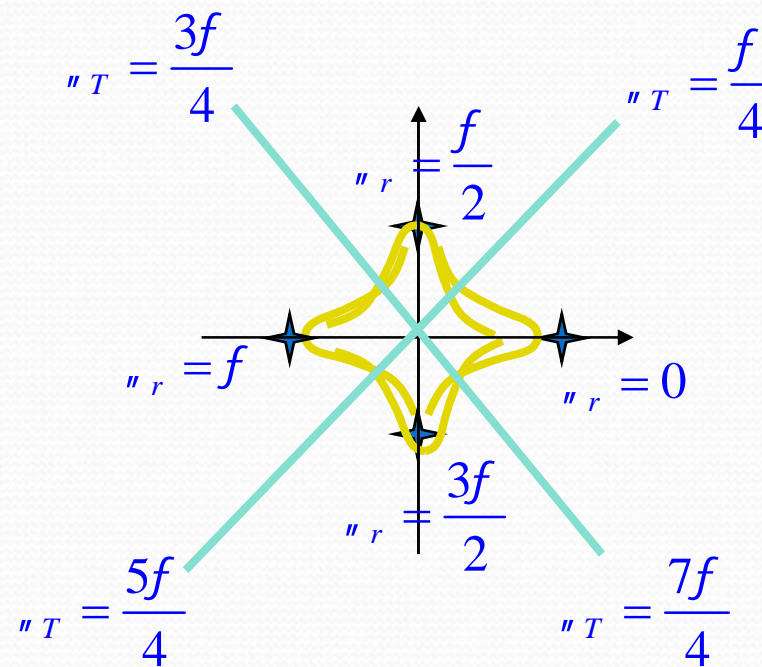
$$r_3 = (n_c, -\sqrt{\mathcal{I}_s} + n_c)$$

What do you think is optimal detector ??

Optimal Detector

- Example of $M=4$
 - Phase of received signal:

$$\theta_r = \tan^{-1}\left(\frac{r_s}{r_c}\right) = \frac{2fm}{M}$$



Probability of Error

- Hard to get closed-form expression

- M=2 case

- Same as binary PAM :

- Where \mathfrak{E}_b is energy per bit

$$P_2 = Q\left(\sqrt{\frac{2\mathfrak{E}_b}{N_0}}\right)$$

- M>4 case

- Approximation:

$$P_M \approx 2Q\left(\sqrt{\frac{2\mathfrak{E}_s}{N_0}} \sin \frac{f}{M}\right) \approx 2Q\left(\sqrt{\frac{2k\mathfrak{E}_b}{N_0}} \sin \frac{f}{M}\right)$$

- Equivalent bit-error probability for M-ary PSK

- Approximation:

$$P_b \approx P_M / k$$

- See figure 7.10 page 293, for Probability of Symbol error for M-ary PSK

DPSK(Differential PSK)

- Differential coding + PSK
- Easy to implement receiver
 - No estimation of carrier phase is required
 - No needs of PLL or Costas loop
- See Chapter 7.0 for details
- See figure 7.14 page 300, for Probability of error
 - Binary DPSK is 3dB poorer than Binary PSK
 - Performance Vs. Easy to implement