

# Properties of the Fourier Transform

# Basic Properties of the Fourier Transform

- **Linearity Property:** Given signals  $x_1(t)$  and  $x_2(t)$  with the Fourier transforms

$$F [x_1(t)] = X_1(f)$$

$$F [x_2(t)] = X_2(f).$$

The Fourier transform of  $\alpha x_1(t) + \beta x_2(t)$  is

$$F [\alpha x_1(t) + \beta x_2(t)] = \alpha X_1(f) + \beta X_2(f).$$

- **Duality Property:**

If  $X(f) = \mathcal{F} [x(t)]$ , then  $x(f) = \mathcal{F} [X(-t)]$  and  $x(-f) = \mathcal{F} [X(t)]$ .

**Proof:**

$$\begin{aligned}\mathcal{F} [X(-t)] &= \int_{-\infty}^{\infty} X(-t)e^{-j2\pi ft} dt \\ &= \int_{-\infty}^{\infty} X(t)e^{j2\pi ft} dt \\ &= x(f).\end{aligned}$$

$$\begin{aligned}\mathcal{F} [X(t)] &= \int_{-\infty}^{\infty} X(t)e^{-j2\pi ft} dt \\ &= \int_{-\infty}^{\infty} X(t)e^{j2\pi(-f)t} dt \\ &= x(-f).\end{aligned}$$



- **Time Shift Property:** A shift of  $t_0$  in the time origin causes a phase shift of  $2\pi f t_0$  in the frequency domain.

**Proof:** 
$$\mathbf{F} [x(t - t_0)] = e^{-j2\pi f t_0} \mathbf{F} [x(t)].$$

$$\mathbf{F} [x(t - t_0)] = \int_{-\infty}^{\infty} x(t - t_0) e^{-j2\pi f t} dt.$$

Let  $t' = t - t_0$

$$\begin{aligned} \mathbf{F} [x(t - t_0)] &= \int_{-\infty}^{\infty} x(t') e^{-j2\pi f (t' + t_0)} dt' \\ &= e^{-j2\pi f t_0} \int_{-\infty}^{\infty} x(t') e^{-j2\pi f t'} dt' \\ &= e^{-j2\pi f t_0} \mathbf{F} [x(t)] = e^{-j2\pi f t_0} X(f). \end{aligned}$$



- **Scaling Property:** For any real  $a \neq 0$ , we have

$$\mathcal{F} [x(at)] = \frac{1}{|a|} X \left( \frac{f}{a} \right).$$

- Proof:  
Case 1:

Let  $a > 0$ ; we have

$$\mathcal{F} [x(at)] = \int_{-\infty}^{\infty} x(at) e^{-j2\pi ft} dt.$$

$$t' = at \quad dt = (1/a) dt'$$

$$\mathcal{F} [x(at)] = \frac{1}{a} \int_{-\infty}^{\infty} x(t') e^{-j2\pi(f/a)t'} dt' = \frac{1}{a} X \left( \frac{f}{a} \right).$$

Case 2:

$$a < 0 \quad \mathcal{F} [x(at)] = \int_{-\infty}^{\infty} x(at) e^{-j2\pi ft} dt.$$

Let  $t' = at$ ; we have

$$dt = (1/a) dt'$$

$$\mathcal{F} [x(at)] = \frac{1}{a} \int_{\infty}^{-\infty} x(t') e^{-j2\pi(f/a)t'} dt' = -\frac{1}{a} X \left( \frac{f}{a} \right).$$

- **Convolution Property:** If the signals  $x(t)$  and  $y(t)$  both possess Fourier transforms, then

Proof:  $\mathcal{F} [x(t) * y(t)] = \mathcal{F} [x(t)] \mathcal{F} [y(t)] = X(f) Y(f)$ .

Convolution

$$x(t) * y(t) = \int_{-\infty}^{\infty} x(\tau) y(t - \tau) d\tau$$

$$\begin{aligned} \mathcal{F} [x(t) * y(t)] &= \int_{-\infty}^{\infty} \left( \int_{-\infty}^{\infty} x(\tau) y(t - \tau) d\tau \right) e^{-j2\pi ft} dt \\ &= \int_{-\infty}^{\infty} x(\tau) \left( \int_{-\infty}^{\infty} y(t - \tau) e^{-j2\pi ft} dt \right) d\tau \\ &= \int_{-\infty}^{\infty} x(\tau) \left( e^{-j2\pi f\tau} Y(f) \right) d\tau \\ &= Y(f) \int_{-\infty}^{\infty} x(\tau) e^{-j2\pi f\tau} d\tau \\ &= X(f) Y(f). \end{aligned}$$



- **Modulation Property:** The Fourier transform of  $x(t) e^{j2\pi f_0 t}$  is  $X(f - f_0)$ , and the Fourier transform of

is  $x(t) \cos(2\pi f_0 t) = x(t) \frac{1}{2} (e^{j2\pi f_0 t} + e^{-j2\pi f_0 t})$

Proof:  $\frac{1}{2} X(f - f_0) + \frac{1}{2} X(f + f_0).$

$$\begin{aligned} \mathcal{F} [x(t) e^{j2\pi f_0 t}] &= \int_{-\infty}^{\infty} x(t) e^{j2\pi f_0 t} e^{-j2\pi f t} dt \\ &= \int_{-\infty}^{\infty} x(t) e^{-j2\pi (f - f_0) t} dt \\ &= X(f - f_0). \end{aligned}$$



- **Parseval's Property:** If the Fourier transforms of  $x(t)$  and  $y(t)$  are denoted by  $X(f)$  and  $Y(f)$ , respectively, then

$$\int_{-\infty}^{\infty} x(t) y^*(t) dt = \int_{-\infty}^{\infty} X(f) Y^*(f) df.$$

- proof:

$$\begin{aligned}\int_{-\infty}^{\infty} x(t) y^*(t) dt &= \int_{-\infty}^{\infty} \left( \int_{-\infty}^{\infty} X(u) e^{j2\pi ut} du \right) \left( \int_{-\infty}^{\infty} Y(v) e^{j2\pi vt} dv \right)^* dt \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} X(u) Y^*(v) e^{j2\pi ut} e^{-j2\pi vt} dt dv du \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} X(u) Y^*(v) \left( \int_{-\infty}^{\infty} e^{j2\pi(u-v)t} dt \right) dv du \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} X(u) Y^*(v) \delta(u-v) dv du \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} X(u) Y^*(u) \delta(u-v) dv du \\ &= \int_{-\infty}^{\infty} X(u) Y^*(u) du \\ &= \int_{-\infty}^{\infty} X(f) Y^*(f) df.\end{aligned}$$



- **Rayleigh's Property:** If  $X(f)$  is the Fourier transform of  $x(t)$ , then

Proof: 
$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(f)|^2 df.$$

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} x(t) x^*(t) dt = \int_{-\infty}^{\infty} X(f) X^*(f) df = \int_{-\infty}^{\infty} |X(f)|^2 df.$$

Parseval's Property



- **Autocorrelation Property:** The (time) autocorrelation function of the **aperiodic** signal  $x(t)$  is denoted by  $R_x(\tau)$  and is defined by

$$R_x(\tau) = \int_{-\infty}^{\infty} x(t) x^*(t - \tau) dt.$$

The autocorrelation property states that

$$\mathcal{F} [R_x(\tau)] = |X(f)|^2.$$

- **Differentiation Property:** The Fourier transform of the derivative of a signal can be obtained from the relation

$$\mathcal{F} \left[ \frac{d}{dt} x(t) \right] = j2\pi f X(f).$$

- **Integration Property:** The Fourier transform of the integral of a signal can be determined from the relation

$$\mathbf{F} \left[ \int_{-\infty}^t x(\tau) d\tau \right] = \frac{X(f)}{j2\pi f} + \frac{1}{2} X(0)\delta(f).$$

- **Moments Property:** If  $\mathbf{F} [x(t)] = X(f)$ , then  $\int_{-\infty}^{\infty} t^n x(t) dt$  the  $n$ th moment of  $x(t)$ , can be obtained from the relation

$$\int_{-\infty}^{\infty} t^n x(t) dt = \left( \frac{j}{2\pi} \right)^n \frac{d^n}{df^n} X(f) \Big|_{f=0}.$$

**TABLE 2.1** TABLE OF FOURIER TRANSFORMS

Time Domain ( $x(t)$ )	Frequency Domain ( $X(f)$ )
$\delta(t)$	1
1	$\delta(f)$
$\delta(t - t_0)$	$e^{-j2\pi f t_0}$
$e^{j2\pi f_0 t}$	$\delta(f - f_0)$
$\cos(2\pi f_0 t)$	$\frac{1}{2}\delta(f - f_0) + \frac{1}{2}\delta(f + f_0)$
$\sin(2\pi f_0 t)$	$-\frac{1}{2j}\delta(f + f_0) + \frac{1}{2j}\delta(f - f_0)$
$\Pi(t) = \begin{cases} 1, &  t  < \frac{1}{2} \\ \frac{1}{2}, & t = \pm\frac{1}{2} \\ 0, & \text{otherwise} \end{cases}$	$\text{sinc}(f)$
$\text{sinc}(t)$	$\Pi(f)$
$\Lambda(t) = \begin{cases} t + 1, & -1 \leq t < 0 \\ -t + 1, & 0 \leq t < 1 \\ 0, & \text{otherwise} \end{cases}$	$\text{sinc}^2(f)$
$\text{sinc}^2(t)$	$\Lambda(f)$
$e^{-\alpha t} u_{-1}(t), \alpha > 0$	$\frac{1}{\alpha + j2\pi f}$
$t e^{-\alpha t} u_{-1}(t), \alpha > 0$	$\frac{1}{(\alpha + j2\pi f)^2}$
$e^{-\alpha t }$	$\frac{2\alpha}{\alpha^2 + (2\pi f)^2}$
$e^{-\pi t^2}$	$e^{-\pi f^2}$
$\text{sgn}(t) = \begin{cases} 1, & t > 0 \\ -1, & t < 0 \\ 0, & t = 0 \end{cases}$	$1/(j\pi f)$
$u_{-1}(t)$	$\frac{1}{2}\delta(f) + \frac{1}{j2\pi f}$
$\delta'(t)$	$j2\pi f$
$\delta^{(n)}(t)$	$(j2\pi f)^n$
$\frac{1}{t}$	$-j\pi \text{sgn}(f)$
$\sum_{n=-\infty}^{n=+\infty} \delta(t - nT_0)$	$\frac{1}{T_0} \sum_{n=-\infty}^{n=+\infty} \delta(f - \frac{n}{T_0})$



Table 4: Basic Continuous-Time Fourier Transform Pairs

Signal	Fourier transform	Fourier series coefficients (if periodic)
$\sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$	$2\pi \sum_{k=-\infty}^{+\infty} a_k \delta(\omega - k\omega_0)$	$a_k$
$e^{j\omega_0 t}$	$2\pi \delta(\omega - \omega_0)$	$a_1 = 1$ $a_k = 0$ , otherwise
$\cos \omega_0 t$	$\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$	$a_1 = a_{-1} = \frac{1}{2}$ $a_k = 0$ , otherwise
$\sin \omega_0 t$	$\frac{\pi}{j}[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$	$a_1 = -a_{-1} = \frac{1}{2j}$ $a_k = 0$ , otherwise
$x(t) = 1$	$2\pi \delta(\omega)$	$a_0 = 1$ , $a_k = 0$ , $k \neq 0$ (this is the Fourier series representation for any choice of $T > 0$ )
Periodic square wave $x(t) = \begin{cases} 1, &  t  < T_1 \\ 0, & T_1 <  t  \leq \frac{T}{2} \end{cases}$ and $x(t + T) = x(t)$	$\sum_{k=-\infty}^{+\infty} \frac{2 \sin k\omega_0 T_1}{k} \delta(\omega - k\omega_0)$	$\frac{\omega_0 T_1}{\pi} \operatorname{sinc}\left(\frac{k\omega_0 T_1}{\pi}\right) = \frac{\sin k\omega_0 T_1}{k\pi}$
$\sum_{n=-\infty}^{+\infty} \delta(t - nT)$	$\frac{2\pi}{T} \sum_{k=-\infty}^{+\infty} \delta\left(\omega - \frac{2\pi k}{T}\right)$	$a_k = \frac{1}{T}$ for all $k$
$x(t) \begin{cases} 1, &  t  < T_1 \\ 0, &  t  > T_1 \end{cases}$	$\frac{2 \sin \omega T_1}{\omega}$	—
$\frac{\sin Wt}{\pi t}$	$X(j\omega) = \begin{cases} 1, &  \omega  < W \\ 0, &  \omega  > W \end{cases}$	—
$\delta(t)$	1	—
$u(t)$	$\frac{1}{j\omega} + \pi \delta(\omega)$	—
$\delta(t - t_0)$	$e^{-j\omega t_0}$	—
$e^{-at} u(t)$ , $\Re\{a\} > 0$	$\frac{1}{a + j\omega}$	—
$te^{-at} u(t)$ , $\Re\{a\} > 0$	$\frac{1}{(a + j\omega)^2}$	—
$\frac{t^{n-1}}{(n-1)!} e^{-at} u(t)$ , $\Re\{a\} > 0$	$\frac{1}{(a + j\omega)^n}$	—

Table 3: Properties of the Continuous-Time Fourier Transform

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

Property	Aperiodic Signal	Fourier transform
	$x(t)$	$X(j\omega)$
	$y(t)$	$Y(j\omega)$
Linearity	$ax(t) + by(t)$	$aX(j\omega) + bY(j\omega)$
Time-shifting	$x(t - t_0)$	$e^{-j\omega t_0} X(j\omega)$
Frequency-shifting	$e^{j\omega_0 t} x(t)$	$X(j(\omega - \omega_0))$
Conjugation	$x^*(t)$	$X^*(-j\omega)$
Time-Reversal	$x(-t)$	$X(-j\omega)$
Time- and Frequency-Scaling	$x(at)$	$\frac{1}{ a } X\left(\frac{j\omega}{a}\right)$
Convolution	$x(t) * y(t)$	$X(j\omega)Y(j\omega)$
Multiplication	$x(t)y(t)$	$\frac{1}{2\pi} X(j\omega) * Y(j\omega)$
Differentiation in Time	$\frac{d}{dt}x(t)$	$j\omega X(j\omega)$
Integration	$\int_{-\infty}^t x(t)dt$	$\frac{1}{j\omega} X(j\omega) + \pi X(0)\delta(\omega)$
Differentiation in Frequency	$tx(t)$	$j \frac{d}{d\omega} X(j\omega)$
Conjugate Symmetry for Real Signals	$x(t)$ real	$\begin{cases} X(j\omega) = X^*(-j\omega) \\ \Re\{X(j\omega)\} = \Re\{X(-j\omega)\} \\ \Im\{X(j\omega)\} = -\Im\{X(-j\omega)\} \\  X(j\omega)  =  X(-j\omega)  \\ \angle X(j\omega) = -\angle X(-j\omega) \end{cases}$
Symmetry for Real and Even Signals	$x(t)$ real and even	$X(j\omega)$ real and even
Symmetry for Real and Odd Signals	$x(t)$ real and odd	$X(j\omega)$ purely imaginary and odd
Even-Odd Decomposition for Real Signals	$x_e(t) = \mathcal{E}v\{x(t)\}$ [x(t) real] $x_o(t) = \mathcal{O}d\{x(t)\}$ [x(t) real]	$\Re\{X(j\omega)\}$ $j\Im\{X(j\omega)\}$

Parseval's Relation for Aperiodic Signals

$$\int_{-\infty}^{+\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |X(j\omega)|^2 d\omega$$

# Fourier Transform for Periodic Signals

- Let  $x(t)$  be a periodic signal with period  $T_0$ . Let  $\{x_n\}$  denote the Fourier series coefficients corresponding to this signal. Then

$$x(t) = \sum_{n=-\infty}^{\infty} x_n e^{j2\pi \frac{n}{T_0} t}.$$

- Since

$$\mathcal{F} \left[ e^{j2\pi \frac{n}{T_0} t} \right] = \delta \left( f - \frac{n}{T_0} \right),$$

we obtain

$$X(f) = \sum_{n=-\infty}^{\infty} x_n \delta \left( f - \frac{n}{T_0} \right).$$



- If we define the truncated signal  $x_{T_0}(t)$  as

$$x_{T_0}(t) = \begin{cases} x(t), & -\frac{T_0}{2} < t \leq \frac{T_0}{2} \\ 0, & \text{otherwise.} \end{cases}$$

we may have

$$\begin{aligned} x(t) &= \sum_{n=-\infty}^{\infty} x_{T_0}(t - nT_0) \\ &= x_{T_0}(t) * \sum_{n=-\infty}^{\infty} \delta(t - nT_0). \end{aligned}$$

- By using the convolution theorem, we obtain

$$\begin{aligned} X(f) &= X_{T_0}(f) \left[ \frac{1}{T_0} \sum_{n=-\infty}^{\infty} \delta \left( f - \frac{n}{T_0} \right) \right] \\ &= \frac{1}{T_0} \sum_{n=-\infty}^{\infty} X_{T_0}(f) \delta \left( f - \frac{n}{T_0} \right) \\ &= \frac{1}{T_0} \sum_{n=-\infty}^{\infty} X_{T_0} \left( \frac{n}{T_0} \right) \delta \left( f - \frac{n}{T_0} \right). \end{aligned}$$

Comparing this result with  $X(f) = \sum_{n=-\infty}^{\infty} x_n \delta \left( f - \frac{n}{T_0} \right)$ ,

we conclude

$$x_n = \frac{1}{T_0} X_{T_0} \left( \frac{n}{T_0} \right).$$



- Alternative way to find the Fourier series coefficients. Given the periodic signal  $x(t)$ , we carry out the following steps to find  $a_n$  :

1. Find the truncated signal  $x_{T_0}(t)$  .

2. Determine the Fourier transform  $X_{T_0}(f)$  of the truncated signal.

3. Evaluate the Fourier transform of the truncated signal  $X_{T_0}(f)$  at  $f = \frac{n}{T_0}$  , to obtain the  $n$ th harmonic and multiply by  $\frac{1}{T_0}$  .



- Example 2.2.3: Determine the Fourier series coefficients of the signal  $x(t)$  shown in Figure 2.2.

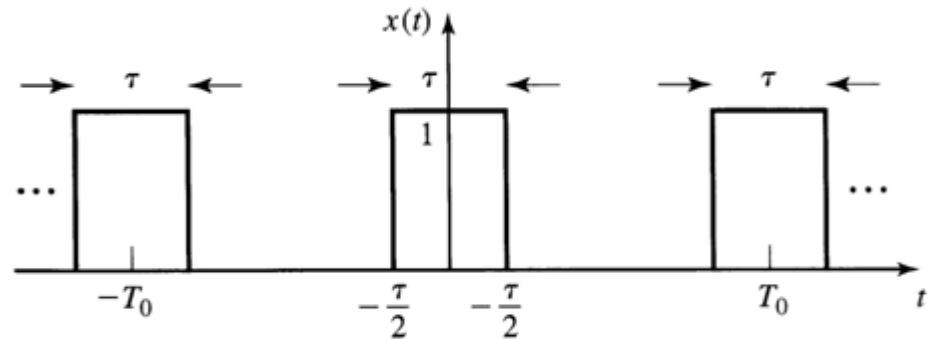


Figure 2.2 Periodic signal  $x(t)$ .

Solution: The truncated signal is

and its Fourier transform is

$$x_{T_0}(t) = \Pi\left(\frac{t}{\tau}\right)$$

Therefore,

$$X_{T_0}(f) = \tau \operatorname{sinc}(\tau f).$$

$$x_n = \frac{\tau}{T_0} \operatorname{sinc}\left(\frac{n\tau}{T_0}\right).$$