

Lecture 24

BJT



Ch6 Basic BJT Amplifiers Circuits

6.3 Frequency Response

Key Words:

Basic Concepts

High-Frequency BJT Model

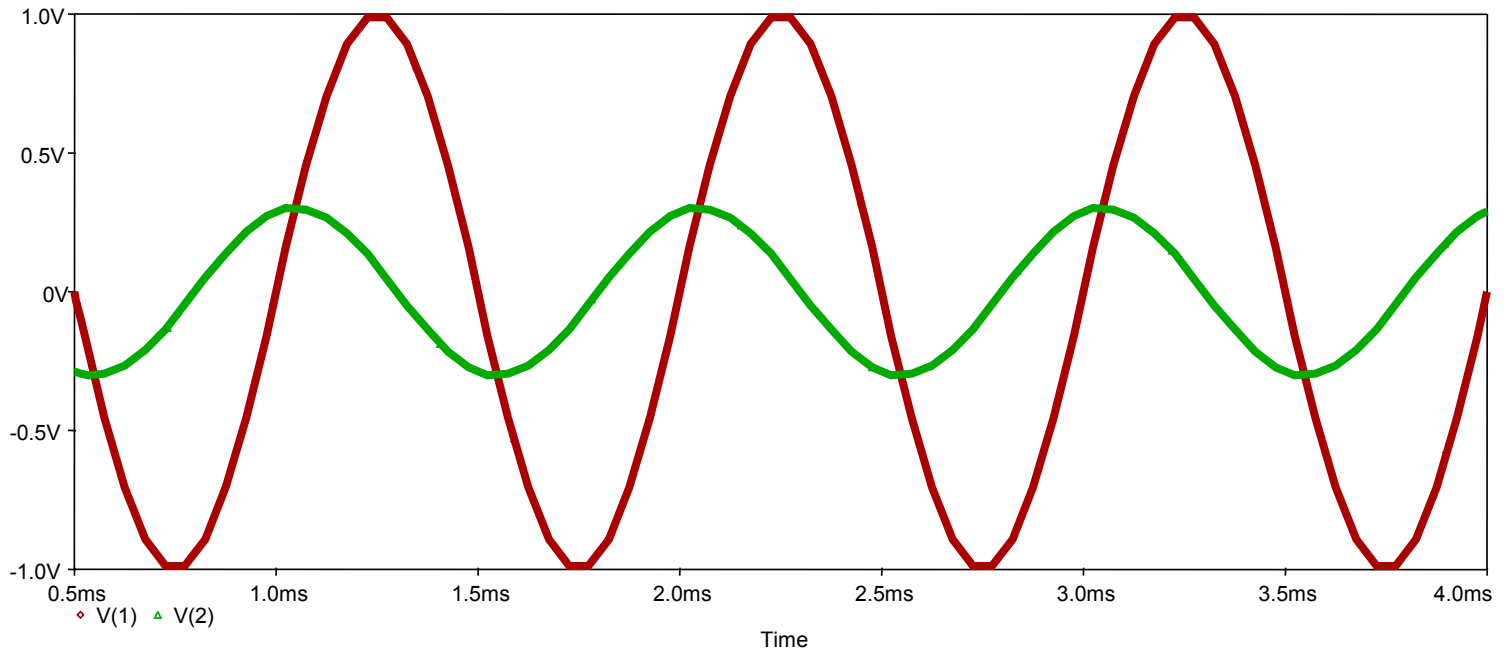
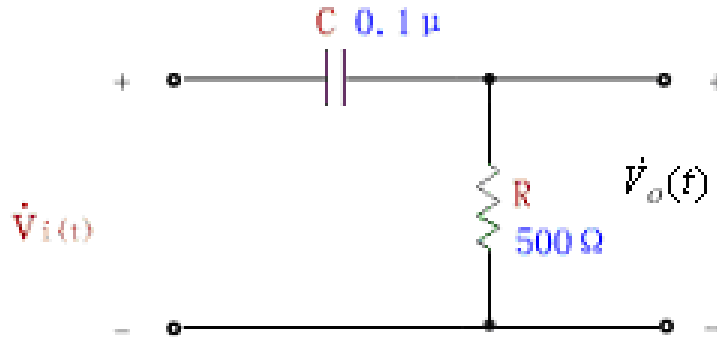
Frequency Response of the CE Amplifier



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Basic Concepts

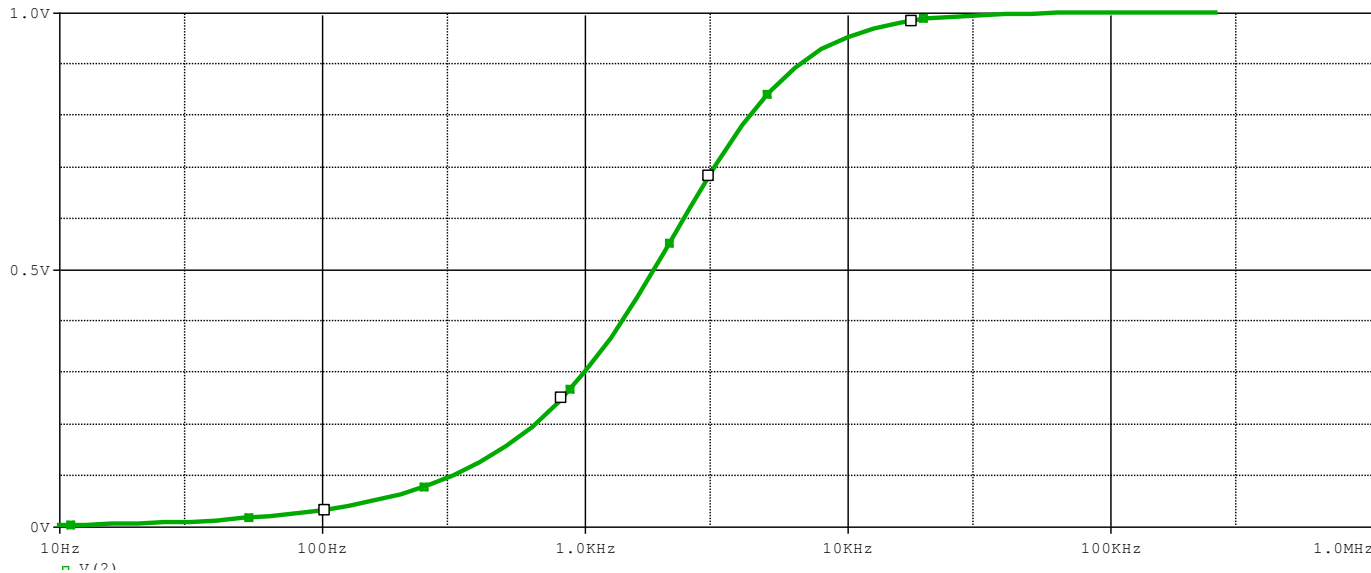
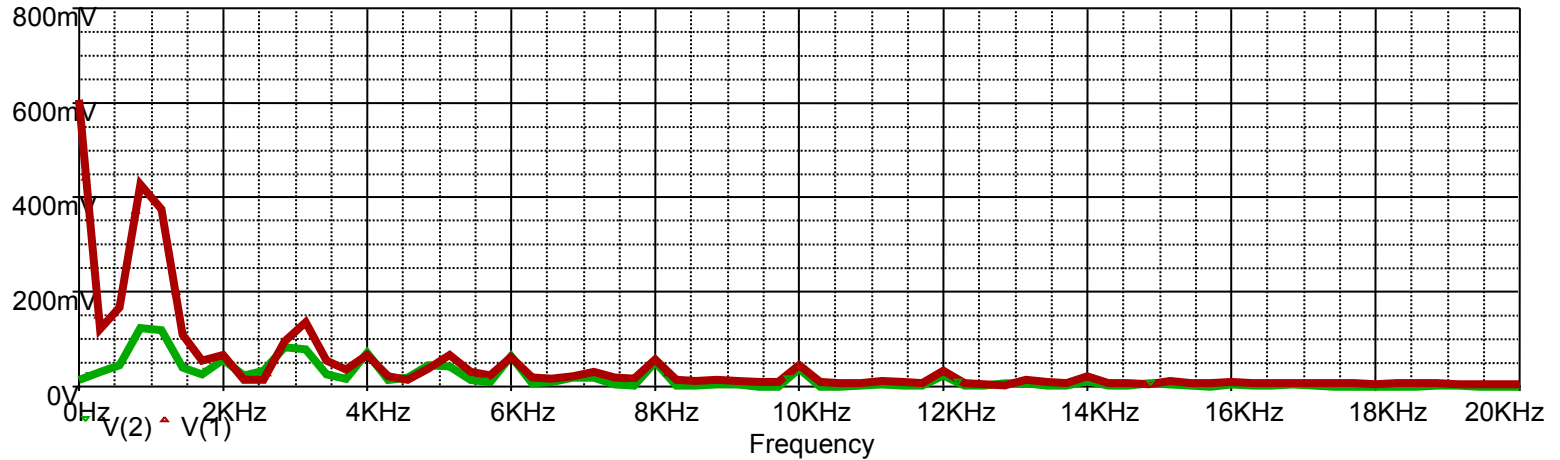




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Basic Concepts

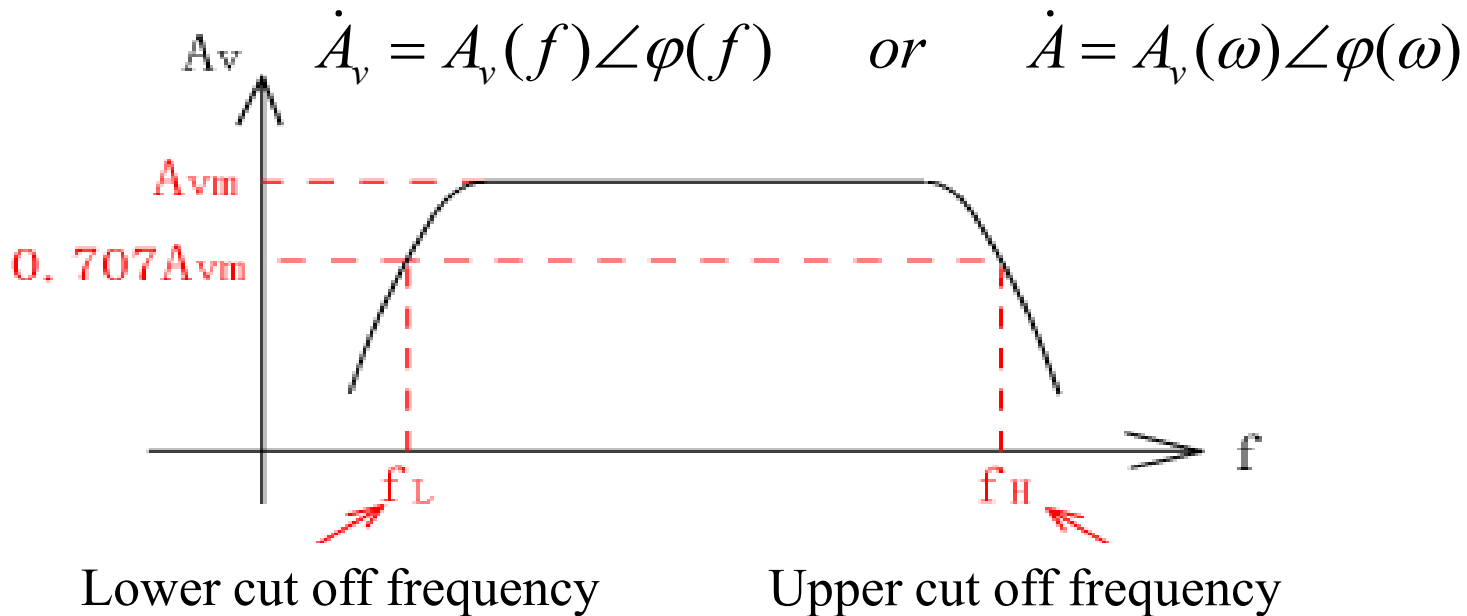




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Basic Concepts



The drops of voltage gain (output/input) is mainly due to:

- 1、 Increasing reactance of C_s, C_c, C_e (at low f)
- 2、 Parasitic capacitive elements of the network (at high f)
- 3、 Disappearance of changing current (for transformer coupled amp.)

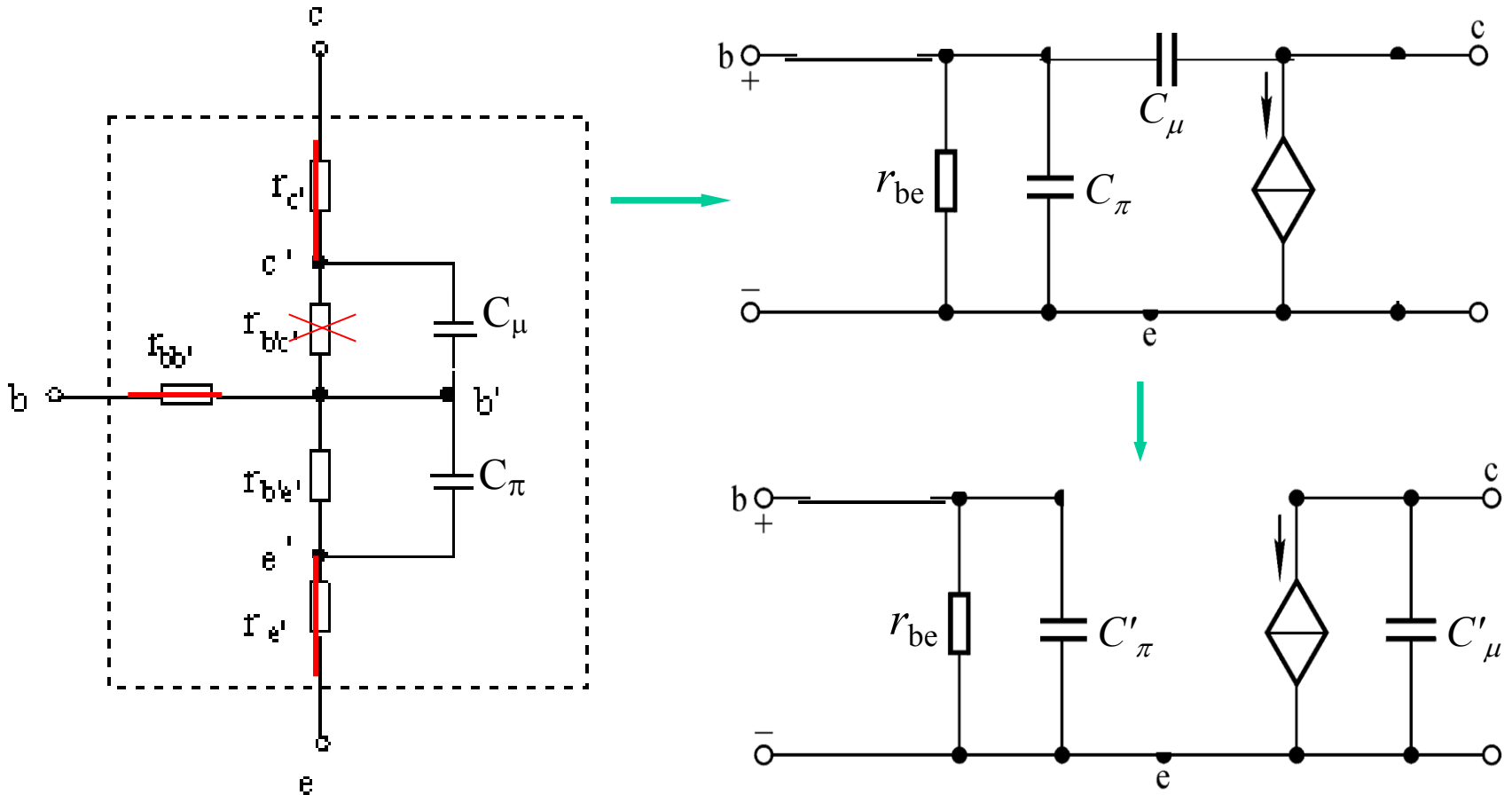


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High-Frequency BJT Model

In BJTs, the PN junctions (EBJ and CBJ) also have capacitances associated with them

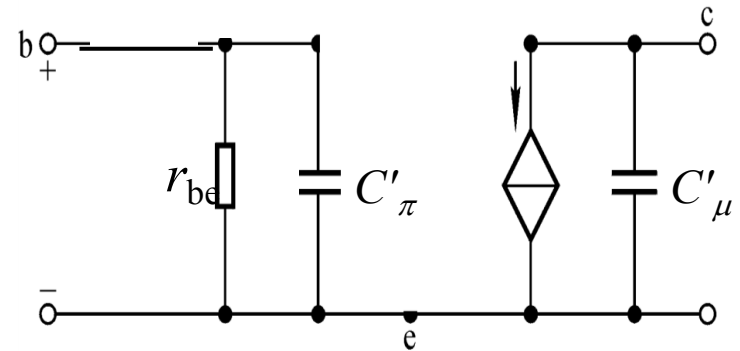
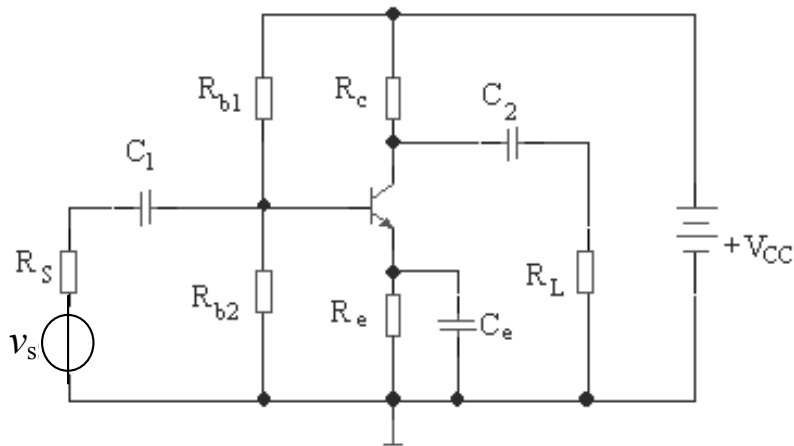




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There are three capacitors in the circuit.

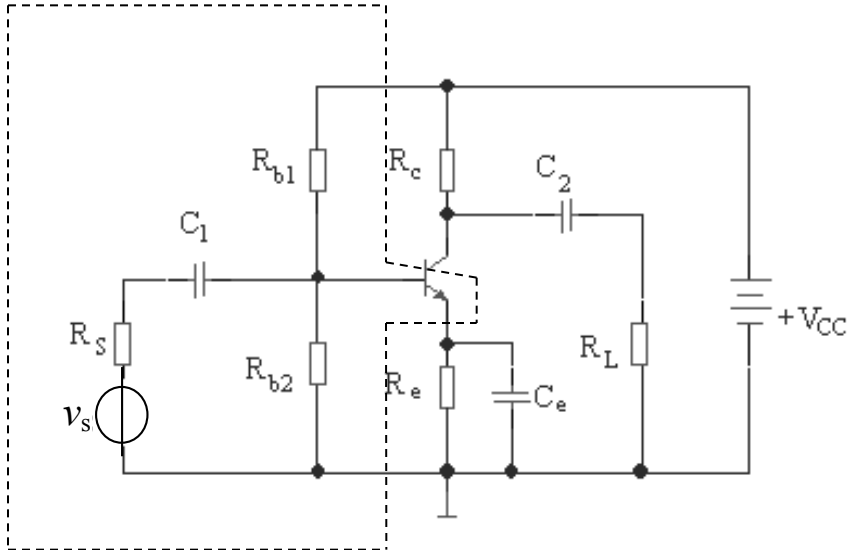
At the mid frequency band, these are considered to be short circuits and internal capacitors C'_π and C'_μ are considered to be open circuits.



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At low frequencies, C_1, C_2 are an open circuit and the gain is zero. Thus C_1 has a high pass effect on the gain, i.e. it affects the lower cutoff frequency of the amplifier.

$$\tau_1 = C_1(R_s + R_{b1} // R_{b2} // r_{be})$$

$$f_{L1} = \frac{1}{2\pi\tau_1}$$

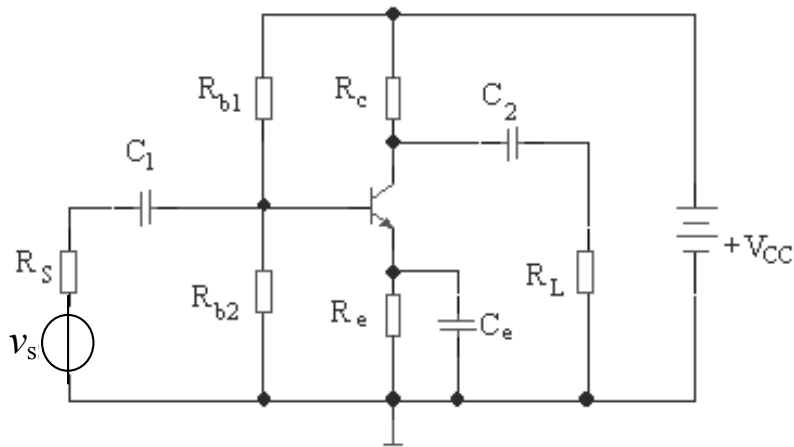
τ_2 is the time constant for C_2 . $\implies \tau_2 \gg \tau_1$ ---is neglected



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$$\tau_1 = C_1(R_s + R_{b1} // R_{b2} // r_{be})$$

$$\tau_2 \gg \tau_1 \quad \text{---is neglected}$$

Capacitor C_e is an open circuit. The pole time constant is given by the resistance multiplied by C_e .

$$\tau_e = \left(\frac{(R_b // R_s + r_{be})}{1 + \beta} // R_e \right) C_e$$

$$f_L \approx 1.1 \sqrt{f_{L1}^2 + f_{L2}^2 + \dots + f_{Le}'^2}$$

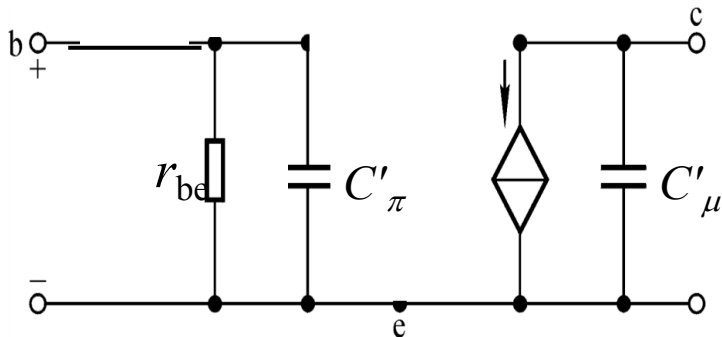
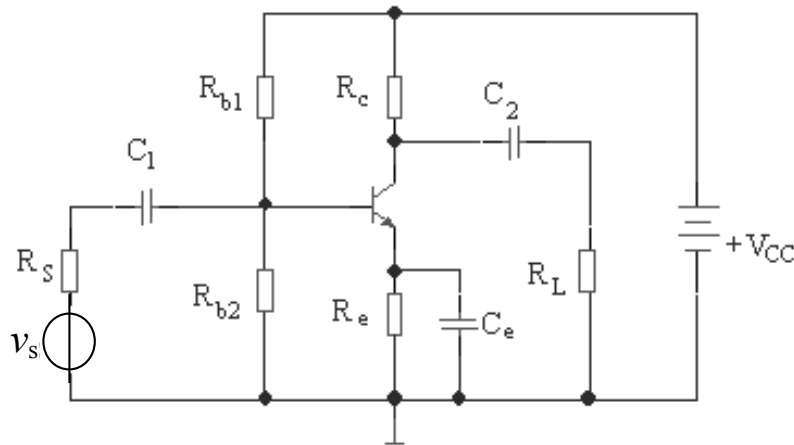
$$f_{Le} = \frac{1}{2\pi\tau_e}$$



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At high frequencies, C_1, C_2, C_e are all short circuit.

The frequency that dominates is the lowest pole frequency.

The time constant is neglected for C'_μ

$$(R'_L \ll 1/j\omega C'_\mu)$$

$$\tau_{C'_\pi} = (R_b \parallel R_s \parallel r_{be}) C'_\pi$$

$$f_H = \frac{1}{2\pi\tau_{C'_\pi}}$$

In summary: the lower cut off frequency is determined by network capacitance.

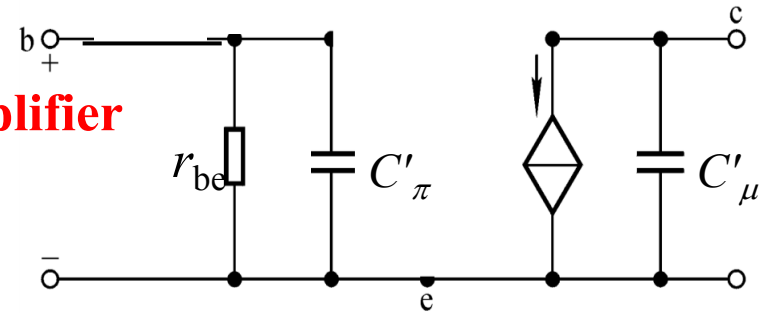
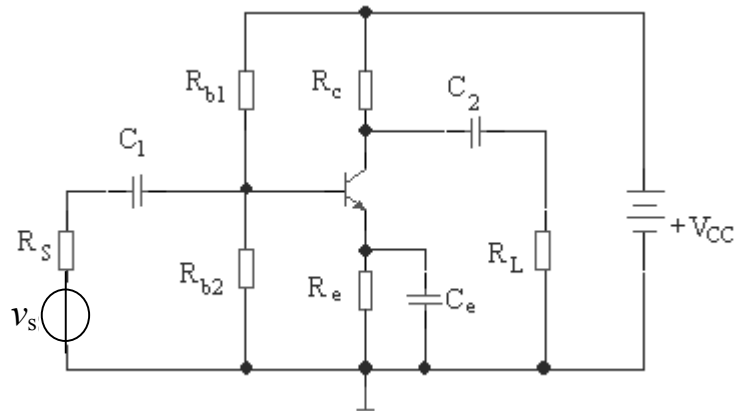
e.g. $C_1(C_2, C_e)$ The higher cut off frequency is determined by the parasitic frequency of the BJT. e.g. C'_π



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$$\dot{A}_v = A_{vm} \cdot \frac{j \frac{f}{f_L}}{(1 + j \frac{f}{f_L})(1 + j \frac{f}{f_H})}$$

For $f_L \ll f \ll f_H$, $\frac{f}{f_L} \rightarrow \infty$, $\frac{f}{f_H} \rightarrow 0 \Rightarrow \dot{A}_v = A_{vm}$ — mid - frequency

For $f < f_L$ ($f \ll f_H$), $\frac{f}{f_H} \rightarrow 0, \Rightarrow \dot{A}_v = A_{vm} \frac{j \frac{f}{f_L}}{1 + j \frac{f}{f_L}}$ — low - frequency

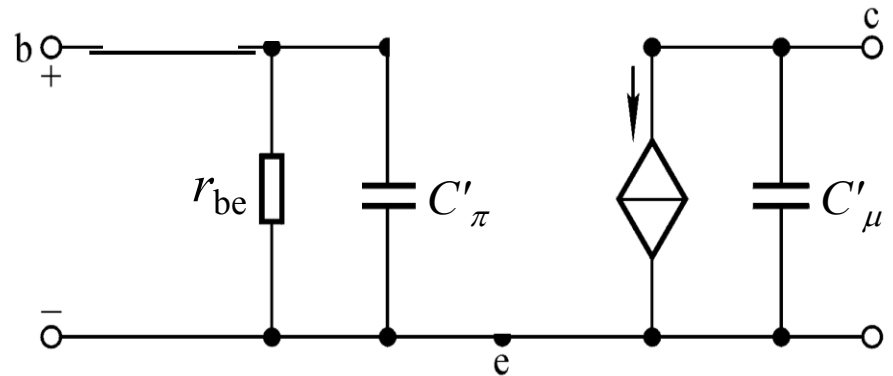
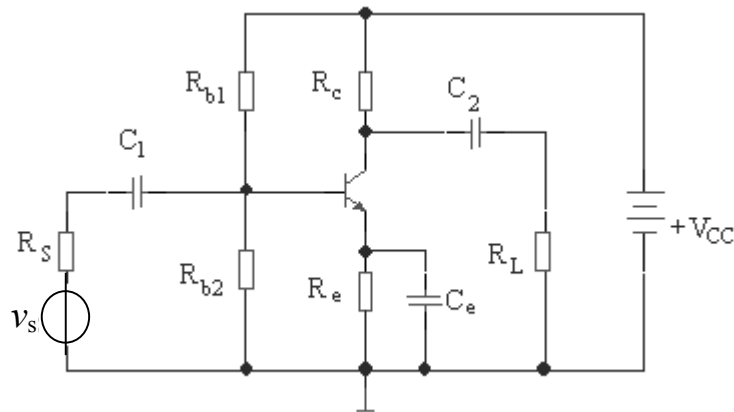
For $f > f_H$ ($f \gg f_L$) $\frac{f_L}{f} \rightarrow 0, \Rightarrow \dot{A}_v = A_{vm} \frac{1}{1 + j \frac{f}{f_H}}$ — High - frequency



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$$A_v = A_{vm} \cdot \frac{j \frac{f}{f_L}}{(1 + j \frac{f}{f_L})(1 + j \frac{f}{f_H})}$$

$$f_L = \frac{\omega_L}{2\pi} = \frac{1}{2\pi\tau_L} \quad f_H = \frac{\omega_H}{2\pi} = \frac{1}{2\pi\tau_H}$$



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