

# Lecture 23

BJT



# Ch6 Basic BJT Amplifiers Circuits

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## 6.3 Frequency Response

**Key Words:**

**Basic Concepts**

**High-Frequency BJT Model**

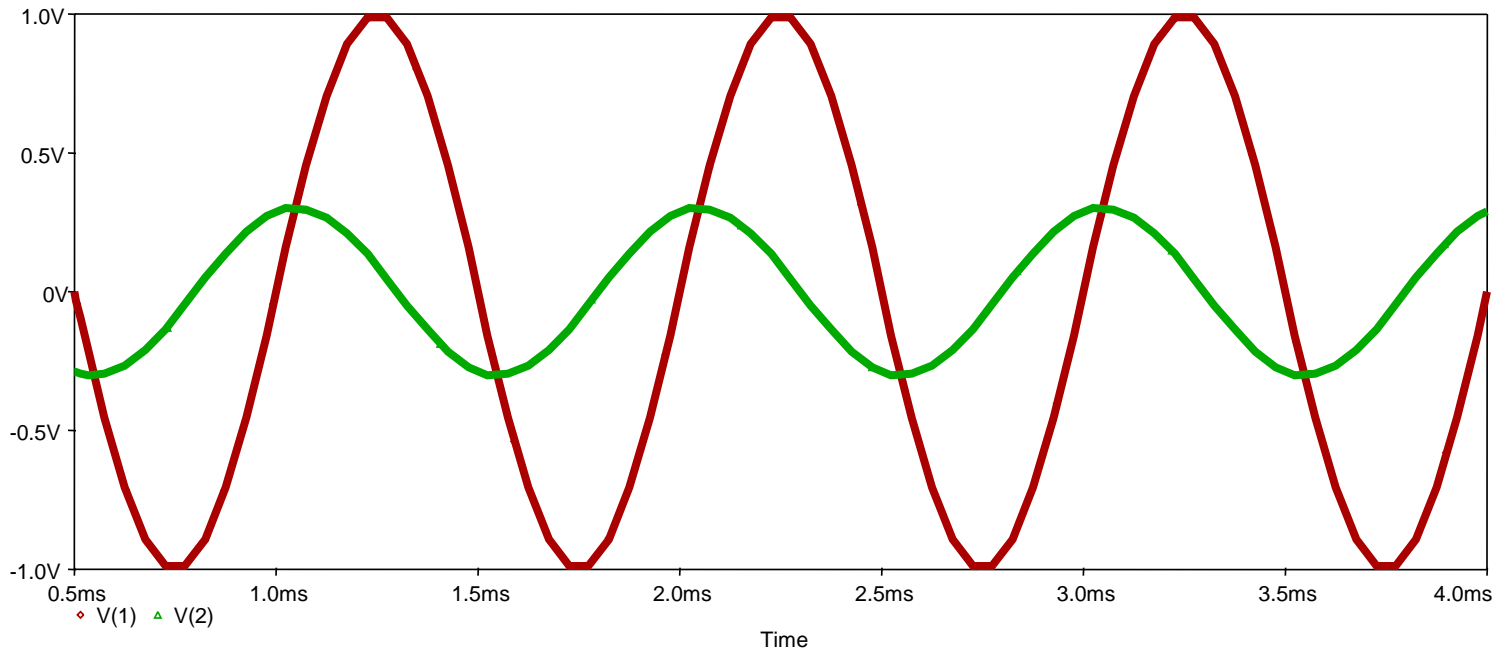
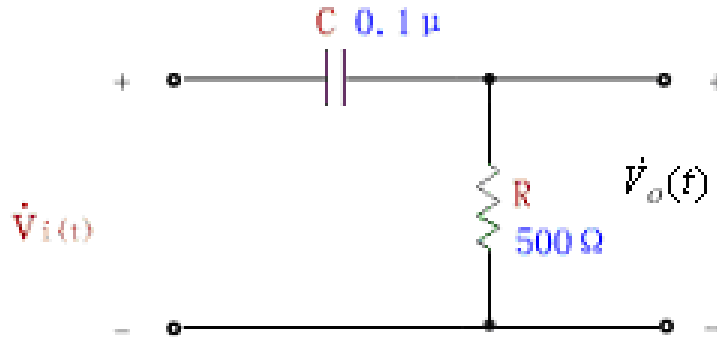
**Frequency Response of the CE Amplifier**



# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Basic Concepts

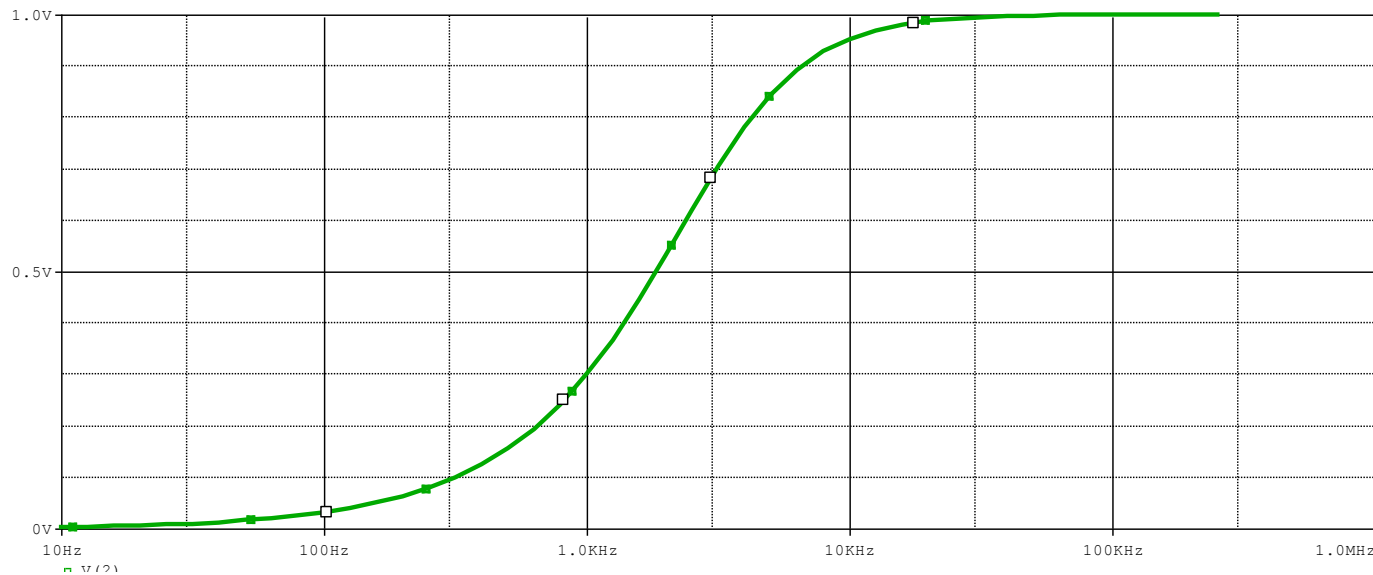
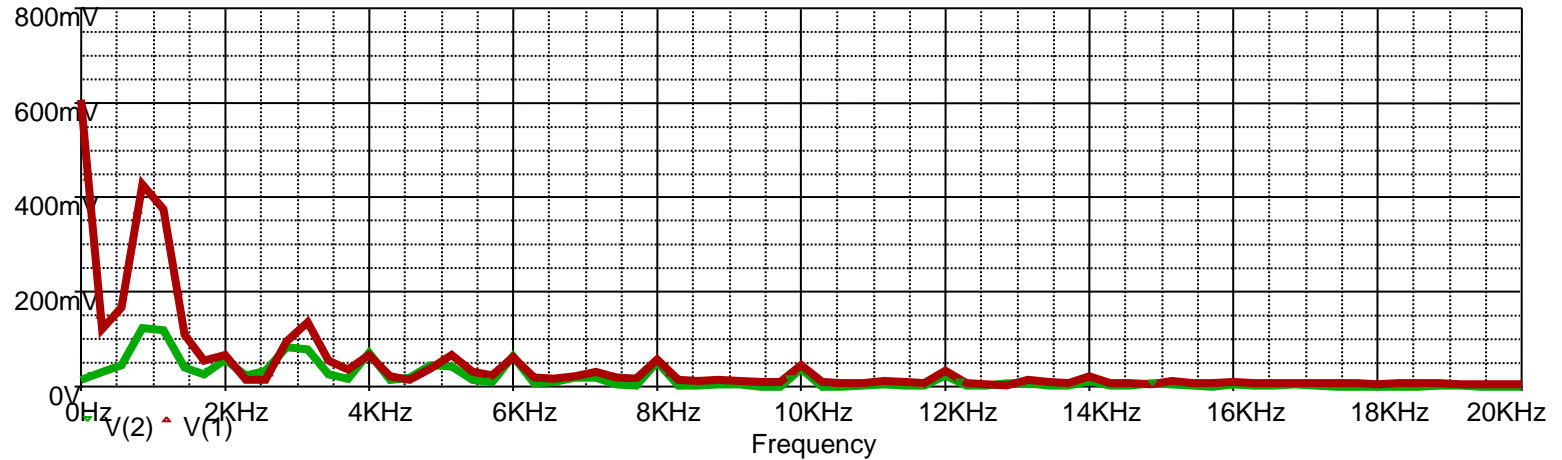




# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Basic Concepts

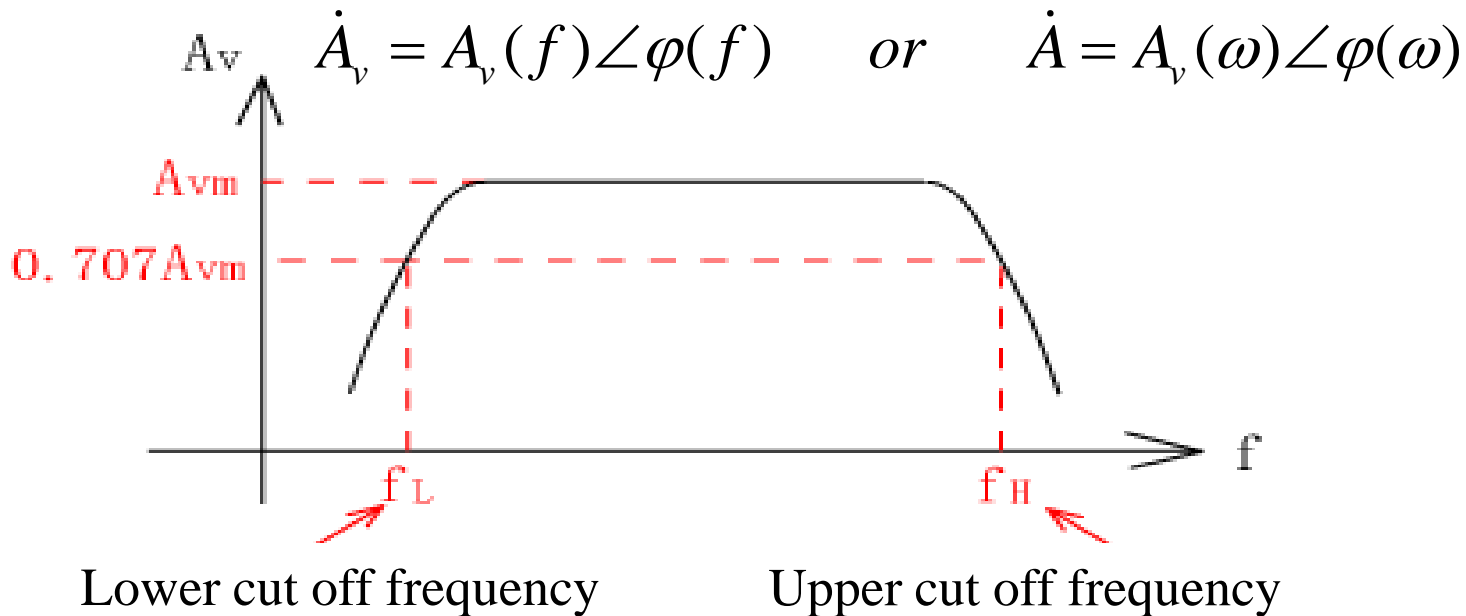




# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Basic Concepts



The drops of voltage gain (output/input) is mainly due to:

- 1、 Increasing reactance of  $C_s, C_c, C_e$  (at low  $f$ )
- 2、 Parasitic capacitive elements of the network (at high  $f$ )
- 3、 Disappearance of changing current (for transformer coupled amp.)

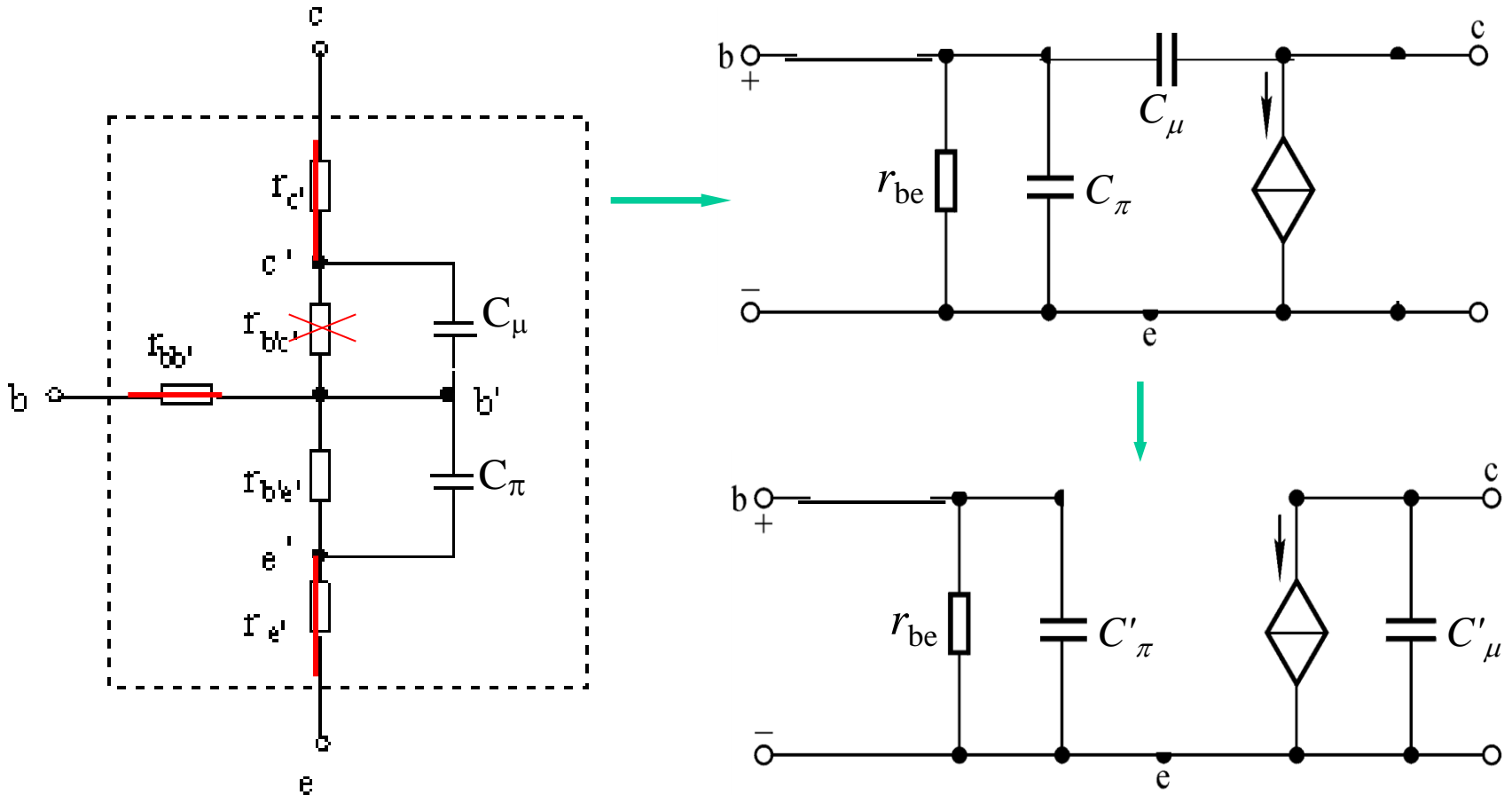


# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### High-Frequency BJT Model

In BJTs, the PN junctions (EBJ and CBJ) also have capacitances associated with them

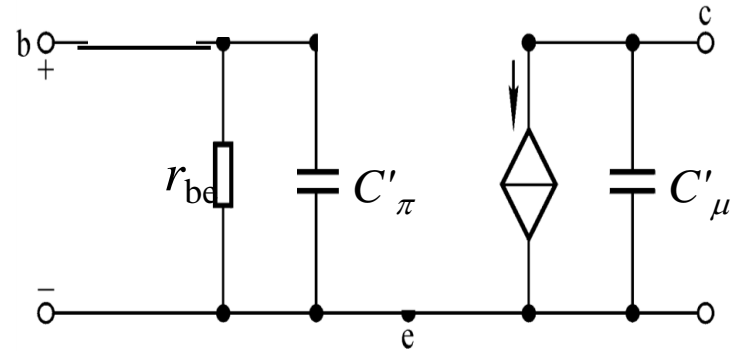
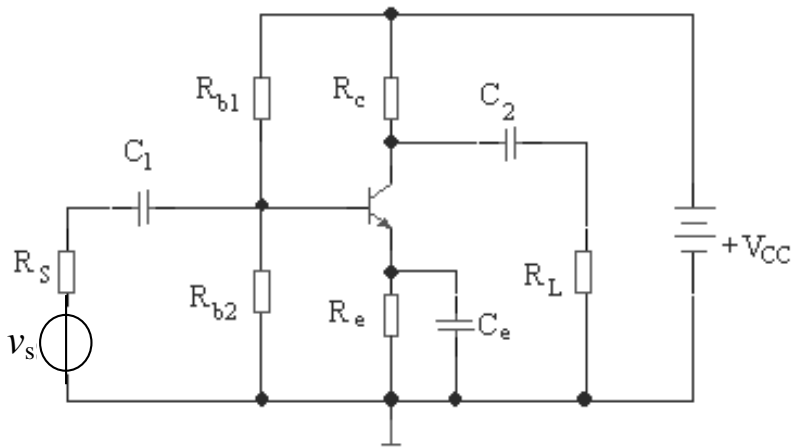




# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Frequency Response of the CE Amplifier



There are three capacitors in the circuit.

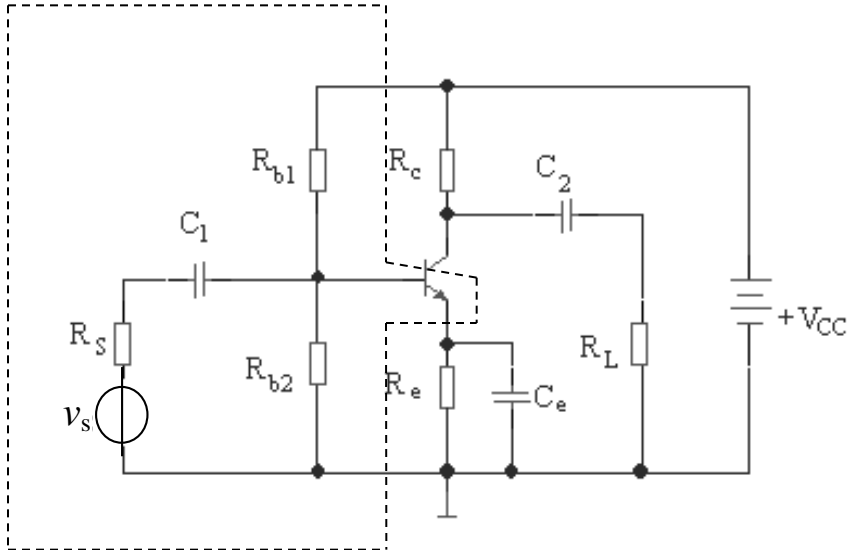
At the mid frequency band, these are considered to be short circuits and internal capacitors  $C'_\pi$  and  $C'_\mu$  are considered to be open circuits.



# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Frequency Response of the CE Amplifier



At low frequencies,  $C_1$ ,  $C_2$  are an open circuit and the gain is zero. Thus  $C_1$  has a high pass effect on the gain, i.e. it affects the lower cutoff frequency of the amplifier.

$$\tau_1 = C_1(R_s + R_{b1} // R_{b2} // r_{be})$$

$$f_{L1} = \frac{1}{2\pi\tau_1}$$

$\tau_2$  is the time constant for  $C_2$ .  $\implies \tau_2 \gg \tau_1$  ---is neglected

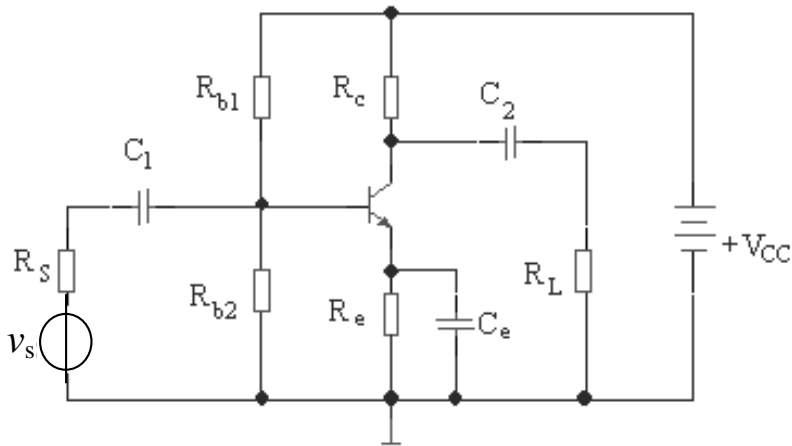




# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Frequency Response of the CE Amplifier



$$\tau_1 = C_1(R_s + R_{b1} // R_{b2} // r_{be})$$

$$\tau_2 \gg \tau_1 \quad \text{---is neglected}$$

Capacitor  $C_e$  is an open circuit. The pole time constant is given by the resistance multiplied by  $C_e$ .

$$\tau_e = \left( \frac{(R_b // R_s + r_{be})}{1 + \beta} // R_e \right) C_e$$

$$f_L \approx 1.1 \sqrt{f_{L1}^2 + f_{L2}^2 + \dots + f_{Le}'^2}$$

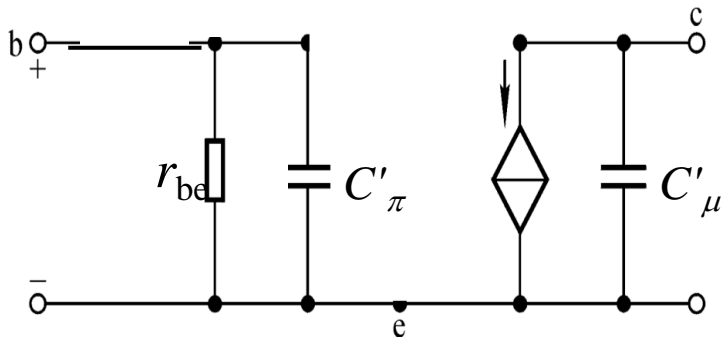
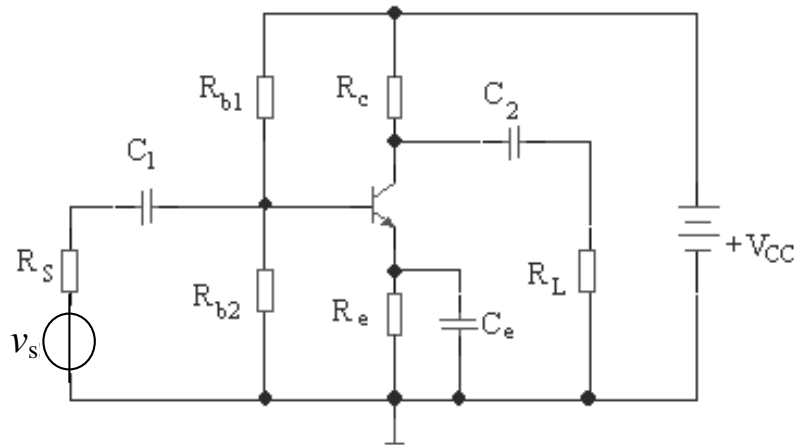
$$f_{Le} = \frac{1}{2\pi\tau_e}$$



# Ch6 Basic BJT Amplifiers Circuits

## 6.3 Frequency Response

### Frequency Response of the CE Amplifier



At high frequencies,  $C_1, C_2, C_e$  are all short circuit.

The frequency that dominates is the lowest pole frequency.

The time constant is neglected for  $C'_\mu$

$$(R'_L \ll 1/j\omega C'_\mu)$$

$$\tau_{C'_\pi} = (R_b \parallel R_s \parallel r_{be}) C'_\pi$$

$$f_H = \frac{1}{2\pi\tau_{C'_\pi}}$$

In summary: the lower cut off frequency is determined by network capacitance.

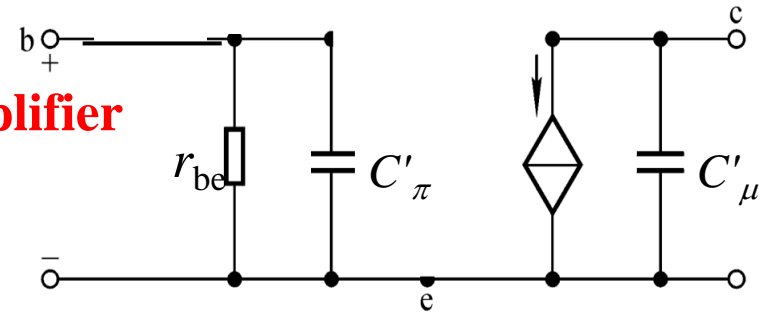
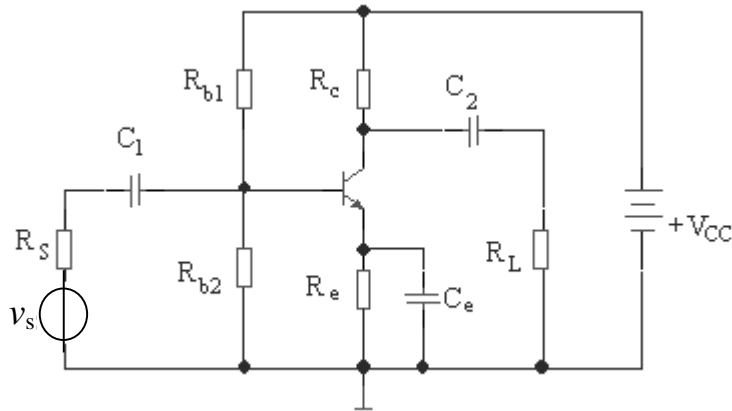
e.g.  $C_1(C_2, C_e)$  The higher cut off frequency is determined by the parasitic frequency of the BJT. e.g.  $C'_\pi$



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### Frequency Response of the CE Amplifier



$$\dot{A}_v = A_{vm} \cdot \frac{j \frac{f}{f_L}}{(1 + j \frac{f}{f_L})(1 + j \frac{f}{f_H})}$$

For  $f_L \ll f \ll f_H$ ,  $\frac{f}{f_L} \rightarrow \infty$ ,  $\frac{f}{f_H} \rightarrow 0 \Rightarrow \dot{A}_v = A_{vm}$  — mid - frequency

For  $f < f_L$  ( $f \ll f_H$ ),  $\frac{f}{f_H} \rightarrow 0, \Rightarrow \dot{A}_v = A_{vm} \frac{j \frac{f}{f_L}}{1 + j \frac{f}{f_L}}$  — low - frequency

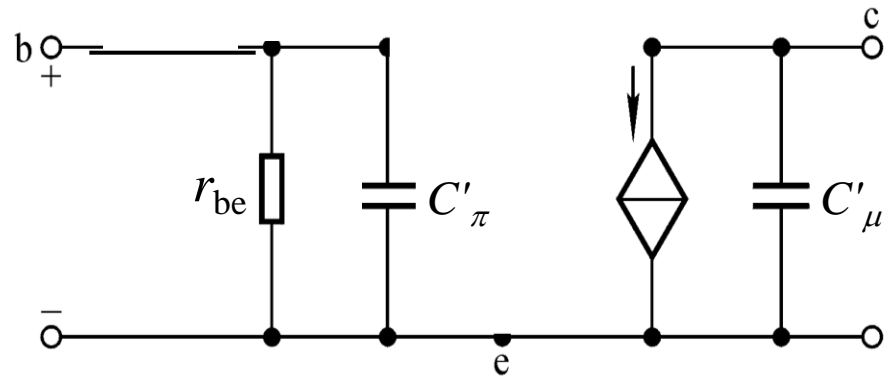
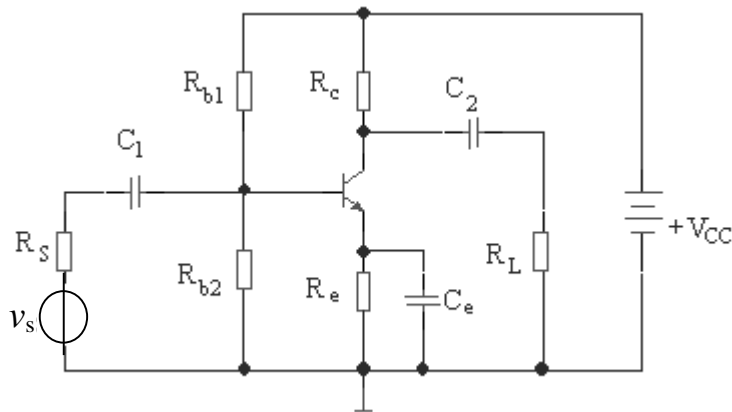
For  $f > f_H$  ( $f \gg f_L$ )  $\frac{f_L}{f} \rightarrow 0, \Rightarrow \dot{A}_v = A_{vm} \frac{1}{1 + j \frac{f}{f_H}}$  — High - frequency



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### Frequency Response of the CE Amplifier



$$A_v = A_{vm} \cdot \frac{j \frac{f}{f_L}}{(1 + j \frac{f}{f_L})(1 + j \frac{f}{f_H})}$$

$$f_L = \frac{\omega_L}{2\pi} = \frac{1}{2\pi\tau_L} \quad f_H = \frac{\omega_H}{2\pi} = \frac{1}{2\pi\tau_H}$$



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### Frequency Response of the CE Amplifier

