



# ELECTRONICS DEVICES AND CIRCUITS

## SECTION - C

## TRANSISTORS

# OBJECTIVE

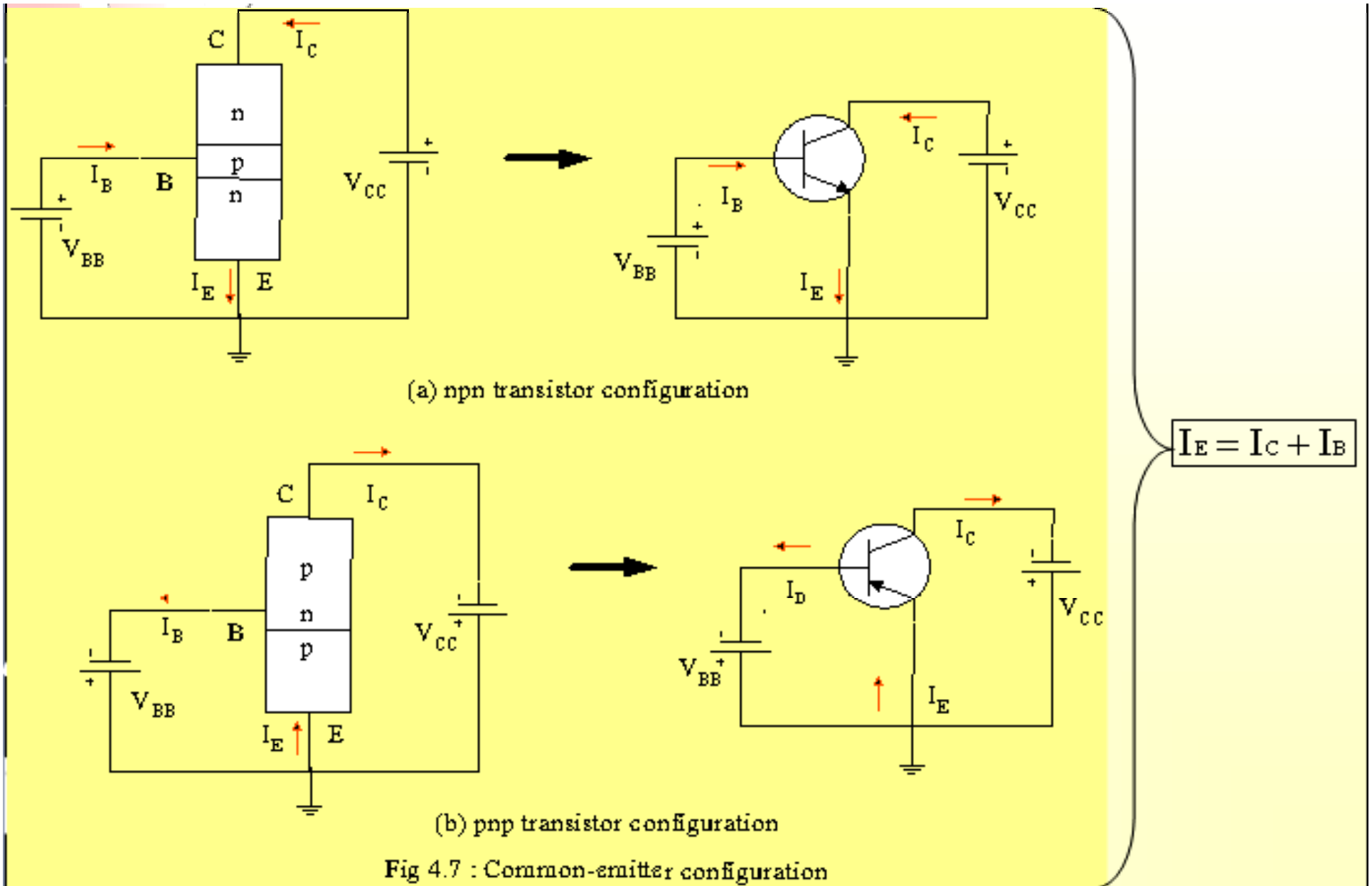
**BJT**

**CE Configuration**

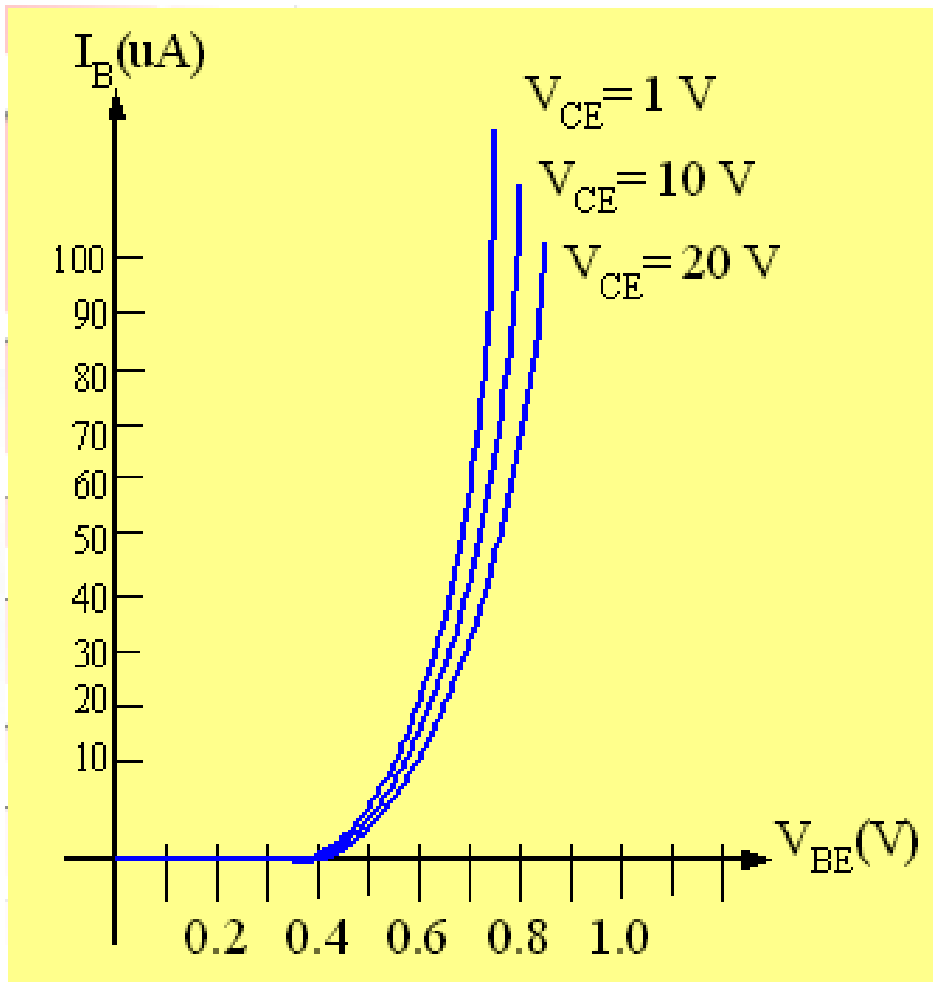
# Common-Emitter Configuration

- It is called common-emitter configuration since :
  - **emitter is common or reference to both input and output terminals.**
  - **emitter is usually the terminal closest to or at ground potential.**
- Almost amplifier design is using connection of CE due to the high gain for current and voltage.
- Two set of characteristics are necessary to describe the behavior for CE ;input (base terminal) and output (collector terminal) parameters.

# Proper Biasing common-emitter configuration in active region

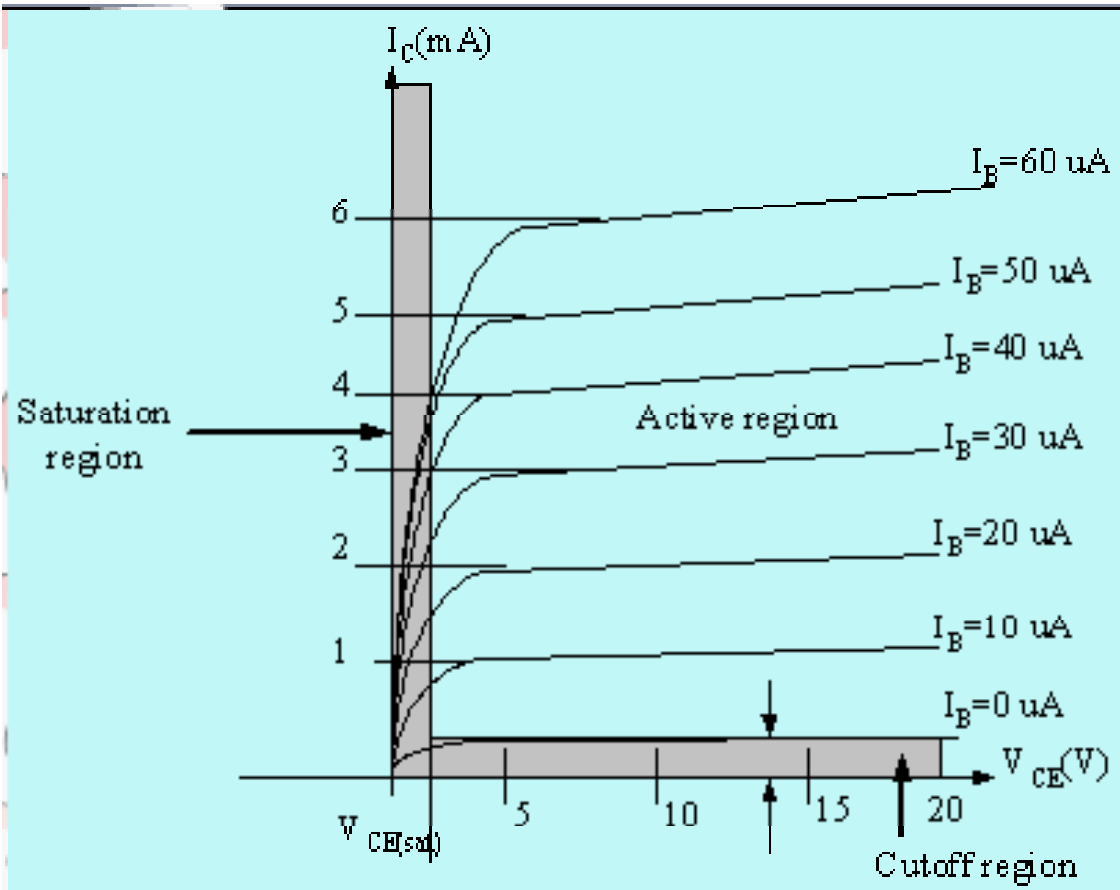






**Input characteristics for a common-emitter NPN transistor**

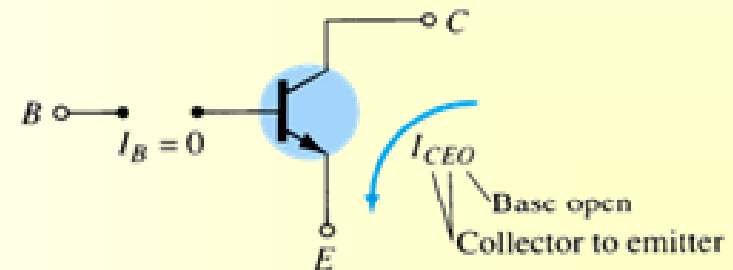
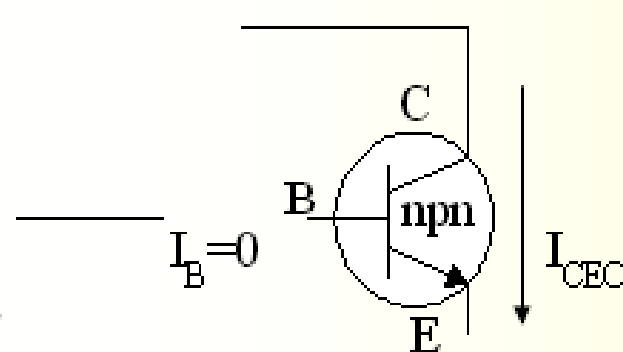
- $I_B$  is microamperes compared to milliamperes of  $I_C$ .
- $I_B$  will flow when  $V_{BE} > 0.7\text{V}$  for silicon and  $0.3\text{V}$  for germanium
- Before this value  $I_B$  is very small and no  $I_B$ .
- Base-emitter junction is forward bias
- Increasing  $V_{CE}$  will reduce  $I_B$  for different values.



Output characteristics for a common-emitter npn transistor

- For small  $V_{CE}$  ( $V_{CE} < V_{CESAT}$ ,  $I_C$  increase linearly with increasing of  $V_{CE}$
- $V_{CE} > V_{CESAT}$   $I_C$  not totally depends on  $V_{CE}$   $\rightarrow$  constant  $I_C$
- $I_B$  ( $\mu\text{A}$ ) is very small compare to  $I_C$  (mA). Small increase in  $I_B$  cause big increase in  $I_C$
- $I_B = 0 \text{ A} \rightarrow I_{CEO}$  occur.
- Noticing the value when  $I_C = 0 \text{ A}$ . There is still some value of current flows.

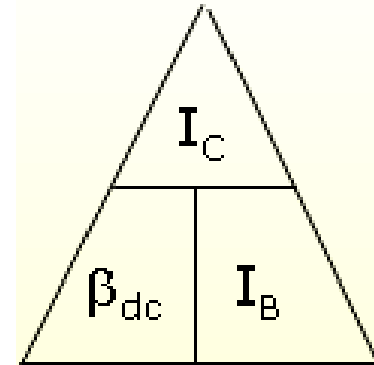
Active region	Saturation region	Cut-off region
<ul style="list-style-type: none"> <li>• B-E junction is forward bias</li> <li>• C-B junction is reverse bias</li> <li>• can be employed for voltage, current and power amplification</li> </ul>	<ul style="list-style-type: none"> <li>• B-E and C-B junction is forward bias, thus the values of <math>I_B</math> and <math>I_C</math> is too big.</li> <li>• The value of <math>V_{CE}</math> is so small.</li> <li>• Suitable region when the transistor as a logic switch.</li> <li>• NOT and avoid this region when the transistor as an amplifier.</li> </ul>	<ul style="list-style-type: none"> <li>• region below <math>I_B=0\mu A</math> is to be avoided if an undistorted o/p signal is required</li> <li>• B-E junction and C-B junction is reverse bias</li> <li>• <math>I_B=0</math>, <math>I_C</math> not zero, during this condition <math>I_C=I_{CEO}</math> where is this current flow when B-E is reverse bias.</li> </ul>



# Beta ( $\beta$ ) or amplification factor

- The ratio of dc collector current ( $I_C$ ) to the dc base current ( $I_B$ ) is dc beta ( $\beta_{dc}$ ) which is dc current gain where  $I_C$  and  $I_B$  are determined at a particular operating point, Q-point (quiescent point).
- It's define by the following equation:

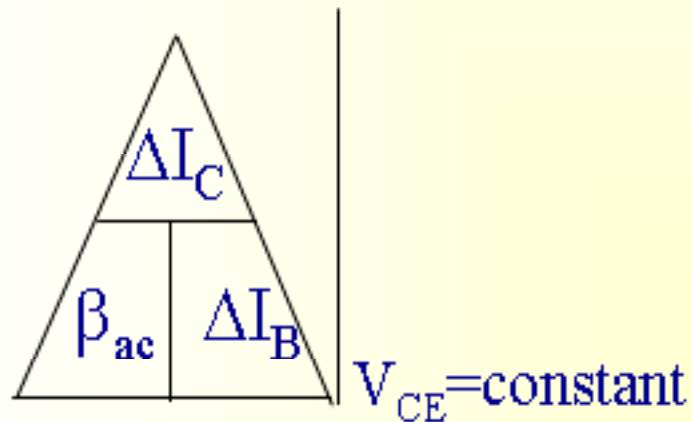
$$30 < \beta_{dc} < 300 \rightarrow 2N3904$$



- On data sheet,  $\beta_{dc} = h_{FE}$  with  $h$  is derived from ac hybrid equivalent cct. FE are derived from forward-current amplification and common-emitter configuration respectively.



- For ac conditions an ac beta has been defined as the changes of collector current ( $I_C$ ) compared to the changes of base current ( $I_B$ ) where  $I_C$  and  $I_B$  are determined at operating point.
- On data sheet,  $\beta_{ac} = h_{fe}$
- It can defined by the following equation:



# Relationship analysis between $\alpha$ and $\beta$

CASE 1

$$I_E = I_C + I_B \quad (1)$$

substitute equ.  $I_C = \beta I_B$  into (1) we get

$$\underline{\underline{I_E = (\beta + 1)I_B}}$$

CASE 2

$$\text{known : } \alpha = \frac{I_C}{I_E} \Rightarrow I_E = \frac{I_C}{\alpha} \quad (2)$$

$$\text{known : } \beta = \frac{I_C}{I_B} \Rightarrow I_B = \frac{I_C}{\beta} \quad (3)$$

substitute (2) and (3) into (1) we get,

$$\underline{\underline{\alpha = \frac{\beta}{\beta + 1}}} \quad \text{and} \quad \underline{\underline{\beta = \frac{\alpha}{1 - \alpha}}}$$